



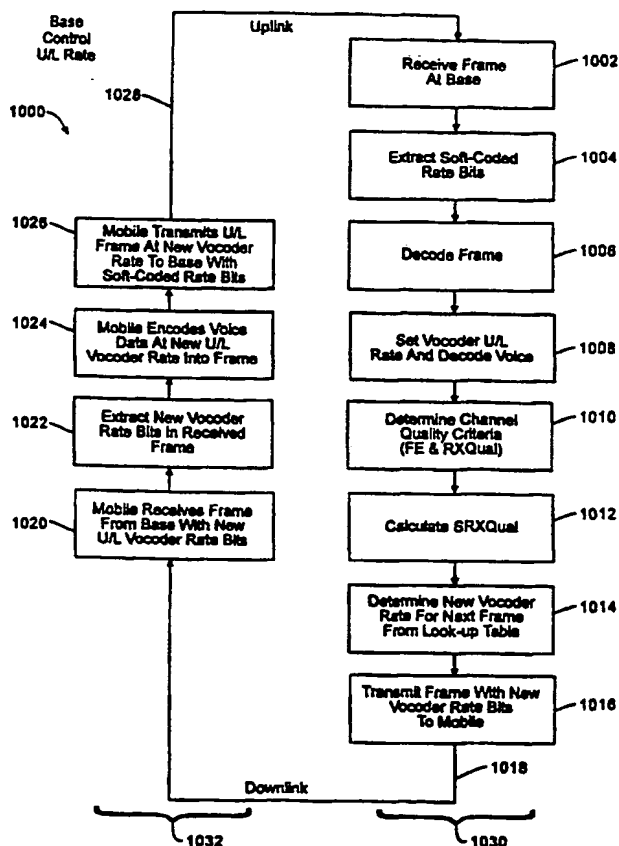
INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification ⁶ : H04L 1/12		A1	(11) International Publication Number: WO 99/60742
			(43) International Publication Date: 25 November 1999 (25.11.99)
(21) International Application Number: PCT/US99/10775		(81) Designated States: JP, European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE).	
(22) International Filing Date: 14 May 1999 (14.05.99)			
(30) Priority Data: 09/080,013 15 May 1998 (15.05.98) US		Published With international search report. Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.	
(71) Applicant: CONEXANT SYSTEMS, INC. [US/US]; 4311 Jamboree Road, Newport Beach, CA 92660-3095 (US).			
(72) Inventors: KOMAILI, Jaleh; 64 Canyon Ridge, Irvine, CA 92612 (US). WAN, Yongbing; 33 Vetrina, Irvine, CA 92606 (US).			
(74) Agent: BENNETT, James, D.; Akin, Gump, Strauss, Hauer & Feld, LLP, Suite 1900, 816 Congress Avenue, Austin, TX 78701 (US).			

(54) Title: SYSTEM AND METHOD FOR RATE ADAPTATION OF A MULTIRATE VOCODER

(57) Abstract

The present invention includes a time-division-multiple-access (TDMA) communication system having a base station and at least one mobile station, each transmitting and receiving an analog radio-frequency signal carrying digital coded speech. The speech is encoded using a vocoder which samples a voice signal at variable encoding rates. During periods when the radio-frequency channel is experiencing high levels of channel interference, the encoded voice channel having a lower encoding rate is chosen. This low-rate encoded voice is combined with the high degree of channel coding necessary to ensure reliable transmission. When the radio-frequency channel is experiencing low levels of channel interference, less channel coding is necessary and the vocoder having a higher encoding rate is used. The high-rate encoded voice is combined with the lower degree of channel coding necessary to ensure reliable transmission. The appropriate levels of channel coding necessary for reliable transmission are determined by various channel metrics, such as frame erasure rate and bit error rate. The determination of the appropriate vocoder rate and level of channel coding for both the uplink and downlink may be determined centrally at the base station, with the vocoder rate and level of channel coding for the uplink being relayed to the mobile station. Alternatively, the appropriate vocoder rate and level of channel coding for the downlink may be determined by the mobile station, and the appropriate vocoder rate and level of channel coding for the uplink may be determined by the base station.



FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BF	Burkina Faso	GR	Greece	ML	Mali	TR	Turkey
BG	Bulgaria	HU	Hungary	MN	Mongolia	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MR	Mauritania	UA	Ukraine
BR	Brazil	IL	Israel	MW	Malawi	UG	Uganda
BY	Belarus	IS	Iceland	MX	Mexico	US	United States of America
CA	Canada	IT	Italy	NE	Niger	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NL	Netherlands	VN	Viet Nam
CG	Congo	KE	Kenya	NO	Norway	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NZ	New Zealand	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	PL	Poland		
CM	Cameroon	KR	Republic of Korea	PT	Portugal		
CN	China	KZ	Kazakhstan	RO	Romania		
CU	Cuba	LC	Saint Lucia	RU	Russian Federation		
CZ	Czech Republic	LI	Liechtenstein	SD	Sudan		
DE	Germany	LK	Sri Lanka	SE	Sweden		
DK	Denmark	LR	Liberia	SG	Singapore		
EE	Estonia						

1

2 SYSTEM AND METHOD FOR RATE ADAPTATION OF A MULTIRATE VOCODER

3 **BACKGROUND OF THE INVENTION**

4 Field of the Invention

5 The present invention relates generally to wireless communication
6 systems. More particularly, the present invention relates to a wireless
7 communication system having an adaptive multi-rate (AMR) vocoder to
8 maximize the voice quality while minimizing the level of channel coding.

9 Description of the Related Art

10 As the use of wireless communication systems become increasingly
11 popular, a variety of methods are being developed to increase the number of
12 mobile communication devices a system can simultaneously service. The
13 Global System for Mobile Communications (GSM), also referred to as the
14 Group Speciale Mobile, is one example of a wireless communication system
15 which is constantly being adapted to increase the number of simultaneous
16 users.

17 The GSM system is modeled after standards created by the European
18 Telecommunications Standards Institute (ETSI) and operates between a
19 telecommunication base station (BS) and a mobile station (MS) using a pair
20 of frequency bands in a frequency division duplex (FDD) configuration. The
21 first frequency band occupies the frequency spectrum between 890 to 915
22 Megahertz (MHZ), and the second frequency band occupies the frequency

1 spectrum between 935 to 960 MHz. Typically, the first frequency range is
2 used for the lower power transmissions from the MS to the BS, and the
3 second frequency range is used for the higher power transmission from the
4 BS to the MS. Each frequency range is divided into 125 channels with 200
5 Kilohertz (kHz) spaced carrier frequencies.

6 The GSM communication system is a time-division-multiple-access
7 (TDMA) system. In the GSM TDMA system, each carrier frequency is divided
8 into eight (8) time slots. Because each MS is assigned a single time slot on
9 one channel in both the first frequency range and the second frequency
10 range, a total of 992 MS may use the BS at the same time.

11 A typical speech channel for GSM communication is sampled at 8 KHz
12 and quantized to a resolution of 13 bits, providing for the digitization of
13 speech ranging from 0 - 4 KHz by a voice encoder, also referred to as a
14 vocoder. The 13 bits are then compressed by a factor of eight (8) in a full-
15 rate vocoder to a voice data digital bit stream of 13 kilobits per second
16 (Kbit/s). Because GSM uses a complex encryption technique with
17 interleaving and convolution coding, a high degree of system integrity and bit
18 error control are achieved. In fact, despite multi-path and co-channel
19 interference, the GSM system may continue to operate despite a carrier-to-
20 interference ratio (C/I) as low as 9 dB, in comparison to a typical advanced
21 mobile phone service (AMPS) analog system requiring a maximum C/I of 17
22 dB.

1 Depending upon the topography of an area, a typical BS may provide
2 communication services to any number of MSs within a radius up to 35
3 Kilometers. Consequently, with the rising popularity of mobile communication
4 devices, it is often the case that during peak periods of use, such as rush-
5 hour traffic, all channels are fully occupied and the BS is not able to provide
6 communication services all of the MS in its region.

7 In order to avoid the inability to service all MS within a region, the ETSI
8 has contemplated a modification of the GSM standard to increase the density
9 of the communication channels. However, because the allocated frequency
10 spectrum of 25 MHZ with 125 separate 200 KHz carrier channels is fixed, a
11 current approach to increasing the density of the communication system is to
12 increase the number of users per channel. In general, this density increase is
13 achieved by decreasing the amount of digital information which is sent to and
14 from each BS, thereby allowing each BS to support more users in a 200 kHz
15 frequency band.

16 One approach to decreasing the amount of digital information passing
17 between a BS and a MS is to decrease the vocoder rate of the digital voice
18 data from a full-rate vocoder rate of 13 kilobits per second (Kbits/s) to a half-
19 rate vocoder rate of 5.6 Kbits/s. Although the ability currently exists to
20 effectively double the number of users on any one communication channel
21 from eight (8) to sixteen (16) by using the half-rate vocoder, it has been found
22 that the 5.6 Kbits/s vocoder rate is barely acceptable as the speech quality is
23 significantly decreased.

1 In light of the above, it would be advantageous to provide a
2 communication system that provides for the user density of a half-rate
3 vocoder system, while providing the voice quality approaching or exceeding
4 that of a full-rate vocoder system. It would also be advantageous to provide
5 a communication system that provides for the modification of the
6 communication channel to incorporate only the amount of channel coding
7 necessary to achieve a reliable communication link between the MS and the
8 BS.

1 **SUMMARY OF THE INVENTION**

2 Broadly, the present invention provides for a wireless communication
3 system having the ability to increase or decrease the vocoder rate and
4 channel coding in response to the level of interference present on the
5 wireless communication channel, resulting in a communication channel
6 having the best possible speech quality. This may be accomplished in either
7 a full-rate or half-rate GSM communication system by decreasing the amount
8 of channel coding during periods of low channel interference to allow
9 transmission of more speech information, representing a higher vocoder rate
10 and resulting in a higher speech quality. During periods of higher channel
11 interference, the amount of channel coding may be increased to the
12 maximum channel coding allowed in a GSM communication network. This
13 increased channel coding provides for consistent and reliable call handling,
14 and results in a lower vocoder rate having a lower speech quality.

15 In an embodiment of the present invention, a time-division-multiple-
16 access (TDMA) communication system includes a base station (BS) and at
17 least one mobile station (MS), each transmitting and receiving an analog
18 radio-frequency signal carrying digitally coded speech. The speech is
19 digitally encoded using a vocoder which samples a voice signal at different
20 encoding rates. Alternatively, the speech may be encoded using a number of
21 different vocoders simultaneously, with each vocoder having a different
22 encoding rate. During periods when the radio-frequency channel is
23 experiencing high levels of channel noise or interference, the encoded voice

1 channel having a lower encoding rate is chosen. This lower-rate encoded
2 voice is combined with the high degree of channel coding necessary to
3 ensure reliable transmission. When the radio-frequency channel is
4 experiencing low levels of channel interference, less channel coding is
5 necessary and the vocoder having a higher encoding rate is used. The high-
6 rate encoded voice is combined with the lower degree of channel coding
7 necessary to ensure reliable transmission. The appropriate level of channel
8 coding necessary for reliable transmission is determined by various channel
9 metrics, such as frame erase rate and bit error rate.

10 The determination of the appropriate vocoder rate and level of channel
11 coding for both the uplink and downlink may be determined centrally at the
12 base station, with the vocoder rate and level of channel coding for the uplink
13 being relayed to the mobile station. Alternatively, the appropriate vocoder
14 rate and level of channel coding for the downlink may be determined by the
15 mobile station, and the appropriate vocoder rate and level of channel coding
16 for the uplink may be determined by the base station.

17 **BRIEF DESCRIPTION OF THE DRAWINGS**

18 The nature, objects, and advantages of the invention will become more
19 apparent to those skilled in the art after considering the following detailed
20 description in connection with the accompanying drawings, in which like
21 reference numerals designate like parts throughout, wherein:

1 Figure 1 is a diagram of a typical wireless telecommunication system,
2 including a base station and a number of mobile stations;

3 Figure 2 is a schematic diagram of the hardware of a typical wireless
4 transceiver of the present invention and includes three separate vocoders,
5 each having a different vocoder rate;

6 Figure 3 is a graph of the relative performance characteristics of a
7 wireless communication system implementing a variable vocoder rate;

8 Figure 4 illustrates the coding, combination and interleaving of speech
9 blocks into a frame, and the variation of the ratio of channel coding to speech
10 coding for various levels of radio-frequency channel noise and interference;

11 Figure 5 is a state diagram illustrating the change of vocoder rate
12 based upon the current status of the communication system, including the FE
13 and BER metrics;

14 Figure 6 depicts a sequence of steps which are performed in the
15 communication system wherein the mobile station calculates the downlink
16 vocoder rate based on its calculations of a number of channel quality metrics;

17 Figure 7 depicts a sequence of steps which are performed in the
18 communication system wherein the mobile station forwards its channel
19 quality metrics to the base station where the vocoder rate for the downlink is
20 determined and communicated to the mobile station;

21 Figure 8 is a quantization table identifying the bits transmitted from the
22 mobile station to the base station in order to provide the base station with the

1 necessary channel metric information to determine the mobile stations

2 vocoder rate;

3 Figure 9 is a quantization table identifying the received bits which

4 correspond to the channel quality metrics made by the mobile station; and

5 Figure 10 depicts a sequence of steps which are performed in the

6 communication system wherein the base station calculates the uplink

7 vocoder rate based on its calculations of a number of channel quality metrics.

1 **DETAILED DESCRIPTION**

2 **System Architecture of a Preferred Embodiment**

3 Referring first to Figure 1, an exemplary communication system of the
4 present invention is shown and generally designated 100. Communication
5 system 100 operates in compliance with the GSM communication standard
6 which includes a time-division-multiple-access (TDMA) communication
7 scheme. In general, a TDMA communication system provides for the
8 transmission of two or more data channels over the same radio-frequency
9 channel by allocating separate time intervals for the transmission of each
10 data channel. In a GSM system, each 200 kilohertz (kHz) radio-frequency
11 channel is divided into repeating time frames, each frame having a duration
12 of 4.615 milliseconds. Each frame contains eight (8) time intervals (also
13 called "slots") each having a duration of 577 microseconds ($4,615/8$) and
14 assigned to a different user.

15 Communication system 100 includes a base station (BS) 102 which
16 receives signals from a mobile switching center (MSC) 106 via
17 communication channel 108. This communication channel includes
18 telephone and/or digital information which may typically originate from land-
19 based telephone systems. Base station 102 transmits information to, and
20 receives information from, mobile stations (MS) 110, 112, and 114 which are
21 within cell 120. Cell 120 is a geographical region within which all mobile
22 stations communicate with the base station 102. Typically, these cells range

1 may have radii ranging from twenty-five (25) to thirty-five (35) kilometers, and
2 may include such geographical disturbances such as buildings 130 or
3 mountains 132. As used herein, the term "information" shall be defined to
4 include digital data, encrypted digital data, convolutionally coded, soft-coded,
5 and/or hard-coded data, digital bits or a bit stream, or any other data type
6 known in the art.

7 Because a GSM-based communication system operates with paired
8 frequency bands in a frequency-division-duplex (FDD) mode, base station
9 (BS) 102 sends information to the mobile station (MS) 110 over a first radio-
10 frequency channel 116, typically in the 890 to 915 MHz range and referred to
11 as the "downlink," and mobile station 110 sends information to the base
12 station 102 over a second radio-frequency channel 118, typically in the 935 to
13 960 MHz range and referred to as the "uplink." Although a GSM-based
14 communication system operates using two frequency bands, it is nonetheless
15 possible to implement the present invention in a system where both the BS
16 and MS transmit and receive over the same radio-frequency channel.

17 Communication system 100 may support a number of mobile stations
18 (MSs) 110, 112, and 114. In fact, under the GSM standard, each 25 MHz
19 frequency band is divided into 125 channels with 200 Kiloherertz (KHz) spaced
20 carrier frequencies. With each carrier frequency supporting eight (8)
21 separate users, a single GSM communication system may support nearly one
22 thousand (1,000) simultaneous users.

1 Given the high possible number of simultaneous users contributing to
2 co-channel interference, and the presence of atmospheric and geographical
3 sources of interference, there are periods of time during which a considerable
4 amount of channel noise and interference is present on the communication
5 system 100. Moreover, the presence of buildings 130 and mountains 132
6 result in multi-path distortion which further degrade the reliability of
7 transmissions through the communication system 100. Additionally, because
8 each MS may be moving in a different direction with respect to the BS, either
9 towards or away from the BS at speeds up to 250 kilometers-per-hour (156
10 miles-per-hour), the possibility that a communication link will be temporarily or
11 permanently disrupted is even higher.

12 In an attempt to minimize the deleterious effects of channel noise and
13 interference on both the uplink and downlink communication channels, a
14 significant amount of channel coding is added to the digital voice data.
15 Channel coding is generally defined to include the process of combining the
16 encoded digital voice data from the vocoder, with any redundant data, parity
17 data, cyclic-redundant-checking (CRC) or other check data necessary to
18 ensure the reliable transmission of the voice data. The code rate is the ratio
19 of data bits to total bits (k/n), and is typically just over one-half ($1/2$) in an
20 ordinary GSM-based system with full-rate vocoders, and just below one-half
21 ($1/2$) in a system with half-rate vocoders.

22 During the channel coding process, the introduction of the necessary
23 error correction, redundant data, parity data, CRC or check data is

1 accomplished by convolutionally coding the digital voice data from the
2 vocoder, with the necessary channel coding data. This results in a
3 convolutionally encoded digital data stream which includes a mixture of voice
4 data and channel coding. As will be more thoroughly discussed in
5 conjunction with Figure 2, this digital data stream is modulated and amplified
6 for transmission over a radio-frequency channel. Upon reception of the
7 modulated digital data stream, the data stream is channel de-modulated and
8 the voice data and channel coding is convolutionally decoded and separated.

9 During periods with high levels of channel noise, the introduction of
10 significant channel coding provides for an increased reliability of the
11 communication channel. On the other hand, a data stream containing a
12 significant amount of channel coding information limits the amount of voice
13 data which can be transmitted, and during periods of low channel noise,
14 results in an inefficient use of the communication channel. Consequently, the
15 present invention monitors the current level of channel noise, and either
16 increases the amount of channel coding to improve channel reliability, or
17 decreases the amount of channel coding to provide for the transmission of
18 more voice data.

19 Although the current GSM-based communication systems dictate a
20 maximum vocoder rate of 13 Kbits/s for a full-rate vocoder system, the
21 present invention contemplates a much higher maximum vocoder rate up to
22 the bandwidth limitation of the wireless communication channel itself. For
23 instance, if the current level of channel noise or interference is minimal, it is

1 possible to provide a communication link having virtually no channel coding
2 and thus providing for a voice data rate of 22 Kbits/s. This would correspond
3 to a voice bandwidth over 4 kHz, resulting in a voice channel having a
4 frequency range and corresponding voice quality well beyond that of a
5 traditional 4 kHz voice bandwidth.

6 Transceiver Architecture

7 Referring now to Figure 2, a circuit diagram of a transceiver of one
8 embodiment of the present invention is shown and generally designated 200.
9 The transmitter portion of circuit 200 includes a microphone element 202,
10 such as an electret-type microphone, that receives an acoustic signal, such
11 as a user's voice, and converts the acoustic voice signal to an analog
12 electrical signal. This analog electrical signal passes through amplifier 204
13 for amplification and filtering, and is supplied to the inputs of three (3)
14 separate voice encoders, or vocoders 206, 208, and 210.

15 A vocoder is an analog-to-digital converter (ADC) which is created
16 especially for the digital encoding and compression of analog voice data.
17 Vocoders are designed around high speed digital signal processors (DSPs)
18 and use a form of linear predictive coding which is intended to model the
19 human vocal chords in order to produce realistic synthetic speech with the
20 minimum of memory. In a GSM communication system with full-rate
21 vocoders, voice data is sampled at the rate of 8 kHz and quantized to a
22 resolution of thirteen (13) bits and compressed to a bit rate of 13 Kbits/s. In a

1 GSM communication system with half-rate vocoders. voice data is sampled at
2 the rate of 8 kHz and quantized to a resolution of thirteen (13) bits and
3 compressed to give a bit rate of 5.6 Kbits/s.

4 In the present invention, vocoders 206, 208, and 210 each receives
5 the amplified voice signal from amplifier 204 and each vocoder is
6 continuously encoding the acoustic voice signals at different rates. For
7 example, vocoder 206 may encode the voice signal at a vocoder rate of 8
8 Kbits/s, vocoder 208 may encode the voice signal at a lower vocoder rate of 6
9 Kbits/s, and vocoder 210 may encode the voice signal at an even lower
10 vocoder rate of 4 Kbits/s. The particular vocoder rates discussed herein are
11 merely exemplary, and it is to be appreciated that a vocoder of virtually any
12 rate may be used, so long as the representative digital data rate is capable of
13 being transmitted over the radio-frequency communication channel.

14 The outputs from vocoders 206, 208 and 210 are fed into switch 212
15 which is controlled by processor 214 having a memory storage 215.
16 Processor 214 in the present embodiment is a microprocessor. However,
17 processor 214 may instead be any conventional single or multi-chipped
18 microprocessor, digital signal processor, microcontroller, or any other suitable
19 digital processing apparatus known in the art. Memory storage 215 in the
20 present embodiment may include an electrically erasable programmable
21 read-only-memory (EEPROM), read-only-memory (ROM), random-access-
22 memory (RAM), diskettes or other magnetic recording media, optical storage
23 media, or any combination thereof. Electronic instructions for controlling the

1 operation of processor 214, in the form of program code, may be stored in
2 memory storage 215.

3 Based upon a predefined selection process, described in greater detail
4 below, processor 214 determines the proper vocoder rate and selects the
5 output of the appropriate vocoder 206, 208 or 210 for passage through switch
6 212 to encoder 216. For example, if the voice signal is to be encoded at a
7 full-rate of 8 Kbits/s, then the output of vocoder 206 would be selected by
8 processor 214 and passed through switch 212. Alternatively, if the voice is to
9 be encoded at the rate of 6 Kbits/s, the output of vocoder 208 would be
10 selected. Encoder 216 receives the digital voice data from the vocoder and
11 adds the level of channel coding corresponding to the vocoder rate selected.

12 Once passed through encoder 216, the now-encoded digital voice
13 data is mixed with the analog output of voltage-controlled-oscillator (VCO)
14 218 to modulate the digital voice data onto a carrier frequency in modulator
15 220. Modulator 220 modulates a gaussian-minimum-shift-key (GMSK) signal
16 on a radio-frequency carrier which is then passed through variable power
17 amplifier 222 and through transmit/receive switch 224 to antenna 226 for
18 transmission. A GMSK signal incorporates gaussian-shaped pulses and is
19 intended improve the resilience of the communication channel to co-channel
20 interference. As an alternative to GMSK, other modulation methods known in
21 the art may be used, such as BPSK, QPSK, or FSK.

1 Control of transmit/receive switch 224 is accomplished by processor
2 214 in a method well known in the art. In single antenna transceivers, it is
3 often necessary to switch the antenna between the transmitting and receiving
4 portions of the circuitry in order to isolate the sensitive receiver electronics
5 from the higher power signal generated by the transmitter.

6 The receiver portion of circuit 200 begins with antenna 226 which
7 receives an analog radio-frequency signal that is passed through
8 transmit/receive switch 224 to intermediate-frequency (IF) amplifier and mixer
9 240. Mixer 240 removes the carrier frequency from the radio-frequency
10 signal and passes the remaining analog signal to an analog-to-digital
11 converter (ADC) 242. ADC 242 converts the received analog signal to a
12 digital signal which is then passed through equalization block 244 where the
13 digital signal may be filtered and the digital bit stream recovered, and to
14 channel decoder 246.

15 As will be discussed in more detail below, processor 214 receives a
16 signal, in the form of rate bits, from channel decoder 246. These rate bits
17 identify the appropriate vocoder rate needed to decode the current voice data
18 encoded in the received signal. Based upon the rate bits, processor 214
19 selects vocoder 250, 252, or 254 using switch 248, and the digital voice data
20 from the decoded radio-frequency channel is passed from channel decoder
21 246, through switch 248, and to the appropriate vocoder 250, 252, or 254.
22 For example, if the digital voice data was encoded at a full-rate of 8 Kbits/s,
23 then processor 214 would operate switch 248 to send the digital voice data to

1 vocoder 250 which, in the present embodiment, decodes at the full-rate of 8
2 Kbits/s. Vocoder 250 decodes the digital information received from channel
3 decoder 246, and re-creates the original analog voice signal which is then
4 passed through amplifier 256 and out speaker 258 to be heard by the user.

5 In an alternative embodiment, vocoders 206, 208 and 210 of circuit
6 200 may be replaced by a single vocoder (shown by dashed lines 270)
7 having multiple encoding rates, or a variable encoding rate. For instance, the
8 single variable-rate vocoder 270 may be capable of encoding acoustic signals
9 from amplifier 204 at rates between 4 Kbits/s, and 8 Kbits/s as determined by
10 processor 214. Similarly, vocoders 250, 252, and 254 may be replaced by a
11 single variable vocoder 272.

12 Although the present invention is discussed in conjunction with a
13 TDMA communication system, it is to be appreciated that the use of a TDMA
14 communication scheme is merely exemplary, and the present invention may
15 be practiced on any number of alternative communications systems, such as
16 code-division-multiple-access (CDMA) and frequency-division-multiple-
17 access (FDMA), for example.

18 Referring now to Figure 3, a graphical representation of system
19 performance is shown and generally designated 300. Graph 300 includes a
20 vertical axis 302 labeled "Voice Quality" and a horizontal axis 304 labeled
21 "Carrier-to-Noise Ratio (C/N)." As discussed herein, the term C/N is
22 considered to include a carrier-to-interference (C/I) portion. In summary,

1 graph 300 represents the performance of communication systems based on
2 the level of channel coding and corresponding vocoder rates. More
3 specifically, three separate curves are shown and each represents the
4 performance of a particular system configuration. For example, graph 306
5 represents the performance of a communication system using a full-rate
6 vocoder, with a minimum level of channel coding. As can be seen, curve
7 306 begins at a higher initial voice quality, but as the C/N decreases, the level
8 of interference due to the lower level of channel coding eventually causes a
9 marked decrease in the voice quality.

10 Similarly, curve 312 represents the performance of a communication
11 system using a mid-rate vocoder with a corresponding mid-level of channel
12 coding. Such a mid-rate vocoder rate could be 6 Kbits/s. Although the initial
13 voice quality shown by curve 312 is maintained for a longer period, it too
14 suffers from the interference caused by the lower level of channel coding.

15 Finally, curve 318 represents the system performance of a
16 communication system using a low-rate vocoder with a corresponding higher
17 level of channel coding. In this case, the high level of channel coding
18 provides for a continuous communication link despite a significant decrease
19 in the C/N, however, the voice quality is lower than either the system shown
20 by curve 306 or 312.

21 In order to maintain the highest level of voice quality possible, despite
22 the decreasing C/N, the present invention changes the voice encoding rate

1 and corresponding level of channel coding in order to maximize the voice
2 quality. For example, in environments where the C/N ratio is high, the system
3 uses the highest possible vocoder rate and lowest possible amount of
4 channel coding. In this situation, because of the low levels of noise and
5 interference on the communication channel, there is little need for heavy
6 channel coding to ensure the communication channel is sustained. However,
7 as the C/N ratio begins to decrease, at the precise instant when curve 306
8 crosses curve 312, shown as intersection 310, the communication system of
9 the present invention changes the vocoder rate and corresponding channel
10 coding to the rate associated with curve 312. In this manner, the highest
11 possible level of voice quality is maintained, even though there is a higher
12 level of channel coding present.

13 Similarly, as the voice quality of the system represented by curve 312
14 drops to the level of the system represented by curve 318, shown at
15 intersection 316, the communication system of the present invention again
16 changes the vocoder rate and corresponding channel coding to the rate
17 associated with curve 318. In this manner, the voice quality for the
18 communication channel is always maximized.

19 In the event the channel noise and interference exceeds the maximum
20 allowable level and results in the voice quality being sufficiently poor so as to
21 pass below the threshold 324, shown at intersection 322, the communication
22 channel is terminated. This terminated communication channel is perceived
23 by the user as a "dropped call." Once terminated, system 100 must be re-

1 initialized and a communication channel must be re-established between the
2 BS 102 and the MS 110.

3 Graph 300 has been divided into three (3) regions 308, 314, and 320,
4 representing the maximized voice quality. A communication channel using
5 the present invention will operate within each of these regions as needed to
6 maximize the voice quality. For example, for a communication channel which
7 is initiated at a vocoder and channel coding rate in region 314 corresponding
8 to curve 312, a momentary decrease in the C/N may cause the system to
9 switch to a vocoder and channel coding rate in region 320 corresponding to
10 curve 318. However, once the C/N returns to its original value, the system
11 will shift back to the vocoder and channel coding rate of region 314
12 corresponding to curve 312. In this manner, system 100 may constantly
13 move between regions 308, 314 and 320 to maximize the voice quality of the
14 communication channel.

15 Graph 300 has been shown to include three (3) separate curves
16 representing three (3) different vocoder rates and corresponding levels of
17 channel coding. However, it should be appreciated that the selection of three
18 (3) vocoder rates is merely exemplary, and virtually any number of vocoder
19 rates may be used in the present invention. Moreover, in a system of the
20 present invention incorporating a vocoder having a variable vocoder rate,
21 virtually any combination of vocoder rate and channel coding may be
22 accomplished within the system limits, ranging from a maximum vocoder rate

1 with no channel coding, to minimum vocoder rate with maximum channel
2 coding.

3 Referring now to Figure 4, a diagrammatic representation of the
4 construction of a GSM communication channel is shown and generally
5 designated 400. Representation 400 includes a series of three (3) speech
6 blocks 402, 404, and 406. Speech block 402 includes a channel coding
7 portion 408 and a voice data portion 410. A speech block represents the
8 digital information which has been generated by the vocoder 206, 208 or 210
9 and channel coder 216 of circuit 200. Accordingly, the digital information
10 within a speech block includes both the voice data and channel coding which
11 has been determined necessary for the reliable transmission of the
12 information. While Figure 4 identifies a channel coding portion 408 and a
13 voice data portion 410 as separate portions of speech block 402, it is to be
14 appreciated that such identification is merely for discussion purposes, and the
15 voice data is actually interleaved with the channel coding to create a data
16 stream having 228 bits.

17 Speech blocks 402, 404 and 406 are each shown having different
18 ratios of channel coding portions and voice data portions. More specifically,
19 speech block 402 is shown having a larger proportion of channel coding 408
20 to a smaller proportion 412 of voice coding 410. Speech block 404, on the
21 other hand, has approximately an even proportion 418 of channel coding 414
22 to voice coding 416. Speech block 406 has a larger proportion 424 of
23 channel coding 420 to voice coding 422. In any case, from comparing

1 speech blocks 402, 404, and 406, it can be seen that the ratios of channel
2 coding to voice coding may change, and even though three separate ratios
3 have been shown in Figure 4, virtually any proportion 412, 418, and 424 may
4 be implement with the present invention.

5 In addition to having a variable quantity of voice data and channel
6 coding, a speech block may also be encoded with a number of rate bits 413.
7 These rate bits 413 represent the particular vocoder rate with which the voice
8 data is encoded. For example, in a communication system where vocoder
9 rates may be varied, rate bits 413 provide the necessary vocoder rate
10 information to successfully decode the voice data. In a preferred
11 embodiment, the rate bits are positioned within the speech block 402, but are
12 not convolutionally encoded with the voice data and channel coding. Rather,
13 the rate bits 413 are "soft-coded" into the speech block 402 such that they
14 can be extracted without the need for convolutionally decoding the speech
15 block. The term "extract" in the present context may include convolutionally
16 decoding, soft-decoding, hard-decoding, or any other manner of retrieving the
17 digital information from the data stream known in the art.

18 The "soft-coding" of the rate bits may be accomplished by placing a
19 series of bits within a particular location of the speech block. For example,
20 rate bits 413 may be placed at bit positions 70, 71, and 72 of speech blocks
21 402, 404 and 406. By positioning the rate bits at consistent locations within
22 each of the speech blocks, it is not necessary to decode the block to
23 determine the value of the rate bits. Instead, the value of the bits in bit

positions 70, 71 and 72 could be determined simply by scanning those bits in the serial bit stream. Additionally, it is possible to place the rate bits in more than one location within each speech block, providing for a measure of error correction. For example, rate bits 413 could occur in three separate locations within speech block 402, allowing the averaging of the bits within the three separate locations in order to provide the best approximation of the rate bits despite any transmissions errors.

In a preferred embodiment, rate bits 413 may represent a three-bit binary value corresponding to eight distinct vocoder rates. Table 1 below identifies such a table of eight distinct vocoder rates based upon three rate bits. As can be seen from Table 1, the rate bits 413 may be assigned any vocoder rate within the vocoder range of the communication system.

<u>Rate Bits</u>	<u>Vocoder Rate (Level)</u>	<u>Vocoder Rate</u>
<u>(Kbits/s)</u>		
000	Level 1	3.0
Kbits/s		
001	Level 2	4.0
Kbits/s		
010	Level 3	5.0
Kbits/s		
011	Level 4	6.0
Kbits/s		

1	100	Level 5	7.0
2	Kbits/s		
3	101	Level 6	8.0
4	Kbits/s		
5	110	Level 7	9.0
6	Kbits/s		
7	111	Level 8	10.0
8	Kbits/s		

TABLE 1 - Rate Bits for Corresponding Vocoder Rates

There are three (3) rate bits identified in Table 1, however, the number of rate bits may vary depending upon the total number of vocoder rates available. For instance, if only two rates are available, a single bit would be needed, with a bit value of "0" indicating one rate, and the bit value of "1" indicating the other rate. Similarly, if only four rates were available, two rate bits would be needed, with the bit values of "00" indicating a first vocoder rate, bit values of "01" indicating a second vocoder rate, bit values of "10" indicating a third vocoder rate, and bit values of "11" indicating a fourth vocoder rate.

Although Table 1 includes a series of eight (8) vocoder rates spaced 1 Kbits/s apart, it is to be appreciated that it is not necessary for the vocoder rates to be evenly distributed. In fact, it would be advantageous for the communication system of the present invention to have a number of vocoder

1 rates within the operating range most frequently experienced by the system.
2 For example, if the communication system noise and interference
3 characteristics indicate that the vocoder rate would typically be 6 Kbits/s, then
4 it might be advantageous to provide several vocoder rates within the 5 to 7
5 Kbits/s region in order to maximize the voice quality. In such an environment,
6 a series of eight (8) vocoder rates might include the following vocoder rates:
7 4.0 Kbits/s, 5.0 Kbits/s, 5.5 Kbits/s, 6.0 Kbits/s, 6.5 Kbits/s, 7.0 Kbits/s, 8.0
8 Kbits/s, and 9.0 Kbits/s. Using these vocoder rates would allow the
9 communication system to adjust the vocoder rate just slightly in order to
10 provide the finest possible voice quality during periods of slight fluctuations in
11 the channel noise and interference levels, while retaining the ability to
12 significantly change the vocoder rate for periods of heavy channel noise and
13 interference.

14 Once the voice data has been encoded with the necessary channel
15 coding, and any rate coding, to form speech blocks 402, 404, and 406, each
16 speech block is divided into four (4) sub-blocks. For example, "A" speech
17 block 402 is split into sub-blocks "A₁" 432, "A₂" 434, "A₃" 436, and "A₄" 438.
18 Likewise, "B" speech block 404 is split into sub-blocks "B₁" 440, "B₂" 442, "B₃"
19 444, and "B₄" 446. and "C" speech block 406 is split into sub-blocks "C₁" 448,
20 "C₂" 450, "C₃" 452, and "C₄" 454. In this manner, the 228 bit data stream in
21 the speech block is broken into four (4) sub-blocks of 57 bits each.

22 Using a combination of sub-blocks, a multi-frame 476 is constructed
23 which includes a continual string of data frames, with each data frame having

1 eight (8) time slots 478, 480 and 482. As shown by mapping lines 462 and
2 464 in Figure 4, frame 478 is constructed from the "A₃" sub-block 436 and the
3 "B₁" sub-block 440. This combination of sub-blocks into frames 478, 480 and
4 482 is called "frame-interleaving" and is intended to create a more robust
5 communication channel.

6 In addition to this frame-interleaving, the even bits within frame 478
7 are comprised of the data bits of the "B₁" sub-block 440, and the odd bits
8 within frame 478 are comprised of data bits of the "A₃" sub-block 436. This
9 even bit/odd bit combination is called "bit-interleaving" and results in the
10 distribution of a single speech block over four contiguous frames. This
11 distribution provides for an improved fault tolerance for the communication
12 system, and in circumstances where the noise level and interference level are
13 high, results in a more resilient communication channel.

14 In addition to the combination of sub-blocks 438 and 442, frame 480 is
15 also encoded with communication system specific coding. For example,
16 using the GSM-based communication system of Figure 1, frame 480 is
17 encoded with three (3) leading "tail bits" 482, a first "encoded voice" bit
18 stream 484 of fifty-seven (57) bits, a single "flag" bit 486, a twenty-eight bit
19 "training sequence" 488, a second "flag" bit 490, a second "encoded voice" bit
20 stream 492 of fifty-seven (57) bits, three (3) trailing "tail bits" 494, and an
21 eight and one-quarter (8¼) bit "guard" period 496. The first and second
22 "encoded voice" bit streams 484 and 492 represent the encoded voice which

1 was present in the "B₁" sub-block 440 and the "A₃" sub-block 436, which
2 included both the voice data, channel coding, and rate bits.

3 Because Doppler shift and multi-path echoes in system 100 can affect
4 the received signal quality, each TDMA frame must include training sequence
5 488, also called training bits. The receiver in system 100 compares these
6 training bits with a known training pattern, and from this deduces the transfer
7 function of the propagation path. An adaptive filter is then created within
8 processor 214 to perform the inverse transfer function, thus canceling any
9 unacceptable distortion. This adaptive filtering is well known in the art, and is
10 thus not discussed in more detail here.

11 Because of the frame-interleaving and bit-interleaving employed in this
12 GSM-based communication system 100, it is not possible to decode the voice
13 information without re-assembling the sub-blocks 432 - 454 from successive
14 frames 478 - 482 in multi-frame 476. Consequently, it is necessary for the
15 digitally encoded voice information to be temporarily stored, such as by
16 temporarily placing the encoded voice information into storage 215 of circuit
17 200. Once a sufficient number of frames has been stored in memory storage
18 215, the sub-blocks are then re-assembled and the voice data is decoded
19 from the re-constructed speech blocks, removing all channel coding, and
20 sent through switch 248 to vocoders 250, 252, and 254.

21 A full-rate GSM-based system would assign each time slot within a
22 frame to a different user. For example, each of the eight (8) time slots within

1 a frame would be assigned to eight (8) different users. In a half-rate GSM-
2 based system, the frame and slot timing remains the same, but instead of a
3 user being assigned a time slot in every frame, the user is assigned a time
4 slot in every other frame.

5 **OPERATION**

6 **Communication Channel Metrics**

7 The operation for the present invention includes the modification of the
8 vocoder rate and level of channel coding to provide the best possible voice
9 quality, while ensuring a reliable communication channel. In order to
10 determine the appropriate level of channel coding necessary to provide
11 reliable communication, a number of channel quality metrics are considered
12 by the present invention. Defined generally, these channel quality metrics
13 include characteristics of the communication channel which may be
14 measured, and by continually measuring these channel quality metrics, an
15 accurate evaluation of the channel quality may be made.

16 One channel metric used to evaluate the quality of the communication
17 channel is the uncoded Bit Error Rate (BER). The uncoded BER of a
18 communication channel is defined as the ratio of the number of bits in a data
19 stream which are improperly demodulated to the total number of bits
20 transmitted. In general, a bit error is caused when the noise power level in a
21 communication system becomes comparable to the energy level in each bit
22 transmitted. Consequently, in a system with a small channel-to-noise ratio

(C/N), bit errors are more likely. Conversely, in a system with a large channel-to-noise ratio, bit errors are less likely. Thus, on a fundamental level, the rate of occurrence of bit errors, or the BER, provides an overall system quality metric.

An additional metric which may be used to evaluate the quality of a communication channel is the RX Quality (RXQ) indicator. The RXQ indicator as generally known in the industry is assigned a value by the network, indicating the quality of the received signal based upon the current BER. Table 2 below includes values for a typical network-determined BER with corresponding RXQ values. This table, however, represents an average received quality, and not an instantaneous RXQ value.

<u>RX Qual</u>	<u>Corresponding Bit Error Rate</u>	<u>Range of Actual BER</u>
<u>(%)</u>		
0	Below 0.2	Below 0.1
1	0.2 to 0.4	0.26 to 0.30
2	0.4 to 0.8	0.51 to 0.64
3	0.8 to 1.6	1.0 to 1.3
4	1.6 to 3.2	1.9 to 2.7
5	3.2 to 6.4	3.8 to 5.4
6	6.4 to 12.8	7.6 to 11.0
7	above 12.8	above 15

Table 2 - GSM Standards for RX Quality Metric

The GSM standards for the RXQ of Table 1 is an average value measured during a predefined period of time. However, because the present invention contemplates an immediate response to a decrease in the RXQ value, it is necessary to determine the RXQ metric on a block-by-block basis. This block-by-block calculation of RXQ', for example, would be made within the MS for the downlink, and within the BS for the uplink.

In the present invention, an RXQ' metric is defined and is dynamically measured by re-encoding the decoded voice data coming out of the convolutional decoder, and comparing them against the received bits. The RXQ' value represents the number of bits different between the received bits and the re-encoded bits per block. The RXQ' consequently provides a combined indication of bit error rate and receiver quality for each block.

Referring briefly to Figures 2 and 4, the determination of the RXQ' metric is accomplished by decoding the voice data from a speech block 402 within a received frame 480, and re-coding the voice data for comparison to the encoded received data. The determination of the RXQ' metric takes place within circuit 200 by receiving a transmitted frame 480 and passing the frame through transmit/receive switch 224 to intermediate frequency (IF) amplifier and mixer 240, through ADC 242 and equalizer 244, to channel decoder 246. In channel decoder 246, the frame 480 is decoded to the original speech block which is then passed to storage 215 for later use.

1 Following storage of the original speech block, all channel coding is removed
2 to recover the original voice data which may also be stored in storage 215, or
3 passed on through switch 248 to vocoders 250, 252 or 254 for conversion to
4 audio.

5 Once the original voice data is recovered from channel decoder 246,
6 the now-decoded voice data is then re-encoded through a convolutional
7 coding process identical to that of the channel encoder 216 to exactly re-
8 create the original coded speech block. This re-encoding may be
9 accomplished using channel decoder 246, or the voice data may be passed
10 through a separate channel coder 247. By comparing the original speech
11 block stored in storage 215 with the newly re-coded speech block from
12 channel coder 247, an estimated bit-error-rate may be determined. For
13 example, by comparing the received speech block with the re-coded speech
14 block, the existence of any error-correction which has taken place within
15 channel decoder will become apparent. Consequently, this dynamic method
16 of error detection is considerably more sensitive than other estimates of the
17 BER, and may be done on a block-by-block basis.

18 An additional metric, SRXQ, is defined as the weighed sum of prior
19 RXQ' measurements. The SRXQ metric is intended to introduce some
20 history into the vocoder rate decision making process based on the receiver
21 quality. In one embodiment, the RXQ' measurements for the prior five (5)
22 blocks are considered in the SRXQ measurement. The prior RXQ'
23 measurements are weighted in accordance with the following equation:

1 $SRXQ = \text{SUM } (2^{K-1})(RXQ'(K+4));$

2 where $K=-4, -3, -2, -1$, and 0 , and where $RXQ'(0)$ is the
3 measured value for the most recent block.

4 An alternative channel quality metric, Frame Erase (FE), may be used
5 to determine the overall quality of the channel. The FE metric represents the
6 number of frames which have been determined to be corrupted, and
7 consequently not used in re-generating the original voice data. In other
8 words, the FE metric represents a count of the number of frames which have
9 been erased per unit time. The decision to erase a frame may be made
10 using a number of criterion. In a present embodiment, the determination to
11 erase a frame is made based on the cyclic-redundancy-checking (CRC), also
12 generally known as a "parity" check. Based on a CRC value which is
13 decoded from the received frame, a frame is either used or discarded,
14 avoiding the use of a frame which may have been improperly decoded or
15 otherwise corrupted.

16 System Operation

17 Referring now to Figure 5, a state diagram is shown and generally
18 designated 500. State diagram 500 represents the changes in vocoder and
19 channel coding rates in response to changes in the communication system
20 environment. For discussion purposes, it is assumed that the communication
21 system is initially experiencing a high carrier-to-noise ratio (C/N), and thus the
22 system is initially in state 502 having a relatively high vocoder rate of 8

1 Kbits/s, with a correspondingly low level of channel coding. In other words,
2 state 502 is used in low-noise environments, such as where the carrier-to-
3 interference ration (C/I) exceeds 19 dB, wherein the majority of digital
4 information with a speech block may be voice data. System 100 will remain
5 in state 502 so long as the FE metric remains at zero (0), as indicated by
6 control path 508. This results in a communication system having a superior
7 voice quality.

8 In the event that a frame is erased resulting in the FE metric becoming
9 non-zero, the BER is computed to determine whether it meets or exceeds a
10 threshold value. In the present embodiment, this threshold value is one
11 percent (1%), meaning that if more than one bit out of a total bit stream of
12 one hundred (100) bits is erroneous, the threshold is met or exceeded. Once
13 the FE metric becomes non-zero and the BER is above the one percent (1%)
14 threshold, the system changes to state 504 via control path 510.

15 State 504 is used in environments exhibiting moderate levels of noise
16 and interference, and combines a mid-range vocoder rate of 6 Kbits/s with a
17 moderate level of channel coding. In the current example, the vocoder and
18 channel coding rate will remain at the mid-range of state 504 so long as the
19 BER is greater-than-or-equal-to one percent (1%), and less than five percent
20 (5%) ($1\% \leq \text{BER} < 5\%$). In this state, typically where the C/I is between 10
21 and 19 dB, the communication channel exhibits a reasonably good voice
22 quality.

1 If after a period of time the communication environment improves and
2 the FE metric returns to zero (0) and the BER becomes less than one percent
3 (1%), the system returns to state 502 via control path 512. On the other
4 hand, in the event the system environment becomes more noisy and the
5 channel-to-noise ratio (C/N) becomes smaller, the FE metric will likely
6 increase. If the FE metric increases to equal or exceed 5, and the BER
7 metric is greater-than-or-equal-to 5 percent (5%), (FE>5 and 5% \leq BER) the
8 system passes to state 506 via control path 516. In this state, a higher
9 degree of channel coding is implemented resulting in a corresponding lower
10 vocoder rate of 4 Kbits/s. According to control path 520, the system will
11 remain in state 506 so long as the BER is greater-than-or-equal-to 5% (5% \leq
12 BER), typically when the C/I is between 4 to 10 dB.

13 When the system is in state 506, and the communication environment
14 improves causing the FE metric to decrease to zero (0) and the BER metric
15 to decrease to less than five percent (5%), then the communication system
16 will change to state 504 according to control path 518, thereby decreasing
17 the level of channel coding and improving the voice quality of the system.

18 In the event the communication system is in state 506 and the FE and
19 BER metrics continue to increase, the communication system may eventually
20 discontinue the communication channel resulting in a "dropped call." In the
21 present embodiment, the communication channel will be discontinued when
22 the FE and BER rates exceed 20 and ten percent (10%), respectively, for
23 example.

1 In order to ensure the proper operation of the system of the present
2 invention, it is necessary that the metrics evaluated for determination of the
3 system control between various states 502, 504, and 506 include a measure
4 of hysteresis. For example, if no hysteresis were to be included between
5 states 502 and 504, it would be possible for the system to oscillate rapidly
6 between the two states, resulting in a vocoder rate and level of channel
7 coding which varies from frame to frame. Although this continual vocoder
8 change is possible with the system of the present invention, it is unnecessary
9 and may result in an inefficient use of system resources.

10 The discussion of the various FE and BER values set forth above is
11 intended as one example of a preferred embodiment having three (3)
12 different vocoder and channel coding rates. The FE and BER values set
13 forth are merely exemplary, and any number of alternative FE and BER
14 values may be chosen and implemented. The threshold values for the FE
15 and BER values may be treated as system parameters, and may change for
16 different vocoders. Also, Figure 5 shows a state diagram with three (3)
17 states, however, any number of states may be created within the present
18 invention.

19 Mobile Station Control of Downlink Rate

20 Referring now to Figure 6, a flow chart representing the operation of
21 the communication system of the present invention is shown and generally
22 designated 600. In general, this configuration includes the MS determining

1 the proper downlink vocoder rate and level of channel coding. Following this
2 determination, the MS then transmits the necessary rate information to the
3 BS.

4 Flow chart 600 begins with first step 602 which includes reception of a
5 radio-frequency frame at the MS. Following receipt of the frame at the MS,
6 the soft-coded rate bits are extracted from the frame data in step 604. In a
7 preferred embodiment of the present invention and as discussed above in
8 conjunction with Figure 4, these soft-coded rate bits may include three (3) bits
9 of rate information that can identify up to eight (8) different vocoder and
10 channel coding rates. The frame data is then convolutionally decoded to yield
11 the original speech block in step 605.

12 Using the appropriate vocoder and channel coding rate information
13 extracted in step 604, the speech block is then decoded to recreate the
14 original voice data in step 606. In this manner, the MS may receive a frame
15 containing voice data encoded with virtually any vocoder rate, and the frame
16 may be successfully decoded to the original voice data because all relevant
17 vocoder rate information is transmitted within the frame in the form of soft-
18 coded bits.

19 In order to provide the best possible voice communication channel, the
20 MS determines the channel quality metrics discussed above, such as FE,
21 BER and RXQ, in step 608. The MS also calculates the SRXQ value in step
22 610 and, based upon the results of the measured and calculated metrics,

1 determines the vocoder rate for optimal voice quality in step 612. In a
2 preferred embodiment of the present invention, the rate bits corresponding to
3 the new vocoder and channel coding rate are determined from a look-up
4 table. Once the vocoder and channel coding rate is determined, the MS
5 transmits a frame with the new downlink vocoder rate convolutionally coded
6 into the frame in step 614. Uplink 626 represents the transmission of a frame
7 from the MS to the BS.

8 In step 616, the BS receives the frame containing the convolutionally-
9 coded downlink vocoder rate for the next downlink transmission. Because it
10 is not necessary to know the downlink vocoder rate in order to decode the
11 uplink transmission, the downlink vocoder rate may be convolutionally
12 encoded instead of soft-coded.

13 In step 618, the BS decodes the received frame from the MS yielding
14 the new downlink vocoder rate bits. These vocoder rate bits are used to
15 determine, using a look-up table or the like, the new downlink vocoder rate.
16 Using that newly determined vocoder rate, the BS encodes the voice data in
17 step 619 for transmission to the MS. In step 620, the BS transmits the frame
18 containing the convolutionally encoded voice data and coded downlink
19 vocoder rate bits to the MS. Downlink 628 represents the transmission of a
20 frame from the BS to the MS.

21 Importantly, each downlink message includes as soft-coded bits the
22 rate information related to the speech block. This is so because there exists

1 a possibility that a frame may become corrupted and no longer readable.
2 This corruption may create a situation wherein the MS may have transmitted
3 a message frame in the uplink changing the downlink vocoder rate, and that
4 frame was not successfully received by the BS. If this occurs, the MS would
5 expect to receive a frame having a new vocoder rate, while the frame actually
6 received would be encoded at the old rate. Additionally, in circumstances
7 involving discontinuous transmissions (DTX), such as when the MS is not
8 transmitting to save battery power, the channel characteristics and
9 corresponding vocoder rate information could change significantly between
10 transmitted frames. Consequently, in order to avoid such mis-communication,
11 each speech block is soft-coded with the rate information necessary to
12 decode the speech block.

13 In a preferred embodiment of the present invention as shown in Figure
14 6, steps within sequence 600 identified by bracket 622 are performed within
15 the MS, and steps within sequence 600 identified by bracket 624 are
16 performed within the BS.

17 In any one cycle of uplink-downlink transmissions shown in Figure 6,
18 both the BS and the MS will inform the other of the appropriate vocoder rates
19 for the transmitted message. For example, in an uplink frame containing
20 convolutionally-coded rate bits for the next downlink frame, soft-coded rate
21 bits will be present which will tell the BS what vocoder rate to use in decoding
22 the uplink frame. Similarly, in a downlink frame containing convolutionally-
23 coded rate bits for the next uplink frame, soft-coded rate bits will be present

1 which will tell the MS what vocoder rate to use in decoding that downlink
2 frame.

3 In the communication system of the present invention, it has been
4 termed that vocoder rate bits which are not convolutionally encoded are "soft-
5 coded" into the speech block, and the vocoder rate bits which are
6 convolutionally encoded are "hard-coded" into the speech block. As an
7 alternative terminology, the vocoder rate information which is convolutionally
8 encoded into a speech block could also be considered an "inside" rate, as the
9 vocoder rate information is within the convolutional coding. Vocoder rate
10 information which is soft-coded into the speech-block is considered an
11 "outside" rate, as the vocoder rate information is outside the convolutional
12 coding.

13 Base Station Control of Downlink

14 Referring now to Figure 7, a flow chart representing the operation of
15 an alternative embodiment of the communication system of the present
16 invention is shown and generally designated 700. In general, this
17 configuration includes the MS monitoring a series of channel metrics and
18 relaying this metric information to the BS for determining the proper downlink
19 vocoder rate and level of channel coding. Following this determination, the
20 BS then transmits the soft-coded rate bits to the MS with the following frame.

21 In first step 702, the MS receives a frame with soft-coded rate bits. In
22 step 703, the MS extracts the soft-coded rate bits from the frame, and using a

1 look-up table or the like, determines the appropriate downlink vocoder rate
2 and level of channel coding. In step 704, using this rate information, the MS
3 decodes the frame, yielding a speech block. In step 706, the vocoders are
4 set to the appropriate rate and this speech block is decoded to re-create the
5 original voice in the speech block.

6 During the decoding process, the MS is determining the channel
7 quality of the communication system. For example, quality metrics such as
8 FE and RXQ may be determined in step 708. Following the determination of
9 FE and RXQ, a quantized vocoder value is determined in step 710 which
10 reflects the current communication channel quality. Referring ahead briefly to
11 Figure 8, a quantization table is shown and generally designated 800.
12 Quantization table 800 includes both the FE metric 802 and the RXQ metric
13 804 which are measured at the MS, and lists a number of non-uniform
14 quantization values for each. These RXQ' values are the mid-range of the
15 transmitted quantization levels, and represent a range of RXQ' metric values.
16 Since the FE and RXQ' are both associated with the receiver performance,
17 the quantization of RXQ' is based on the value of FE to effectively quantize
18 RXQ' into eight levels. By locating the current measured values of both the
19 FE and RXQ on the quantization table, a series of three (3) quantization bits
20 are identified. For instance, for a FE value of 1 and a RXQ value of 25,
21 quantization bits 1-0-0 are selected. Once the quantization bits are selected,
22 in step 712 a frame is transmitted from the MS to the BS with the quantization

1 bits fully encoded in the speech block. Uplink 730 represents the
2 transmission of a frame from the MS to the BS.

3 In step 714, the frame is received at the BS with the quantization bits
4 fully encoded. This frame is decoded in step 716 to yield the original
5 quantization bits. Referring briefly to Figure 9, a quantization table 900 is
6 shown which provides a look-up table to reconstruct the FE and RXQ' values
7 from the received quantization bits. For example, for quantization bits 1-0-0,
8 a FE metric value 904 of "1" and an RXQ metric value 906 of "22." These
9 metrics derived from the quantization bits are then used to calculate the
10 SRXQ metric in step 718. Based upon the quantization bits and the results of
11 the SRXQ calculation, a new vocoder rate is determined in step 720 by the
12 MS. In step 722, voice data for the next speech block is encoded using the
13 new vocoder rate, with the new vocoder rate bits being soft-coded into the
14 speech block resulting in a new frame. This new frame is then transmitted
15 from the BS to the MS in step 724. Downlink 732 represents the
16 transmission of a frame from the BS to the MS.

17 Base Station Control of Uplink

18 In addition to the rate bits which are exchanged between the MS and
19 the BS to govern the downlink vocoder and channel coding rate, the rate bits
20 corresponding to the operation of the uplink are also exchanged. This is
21 accomplished by the BS analyzing similar channel quality metrics which are

1 used to determine the appropriate downlink vocoder rate as discussed in
2 conjunction with Figure 6.

3 Referring now to Figure 10, a flow chart representing the operation of
4 an alternative embodiment of the communication system of the present
5 invention is shown and generally designated 1000. In general, this
6 configuration includes the BS monitoring a series of channel metrics
7 determines the proper uplink vocoder rate and level of channel coding.
8 Following this determination, the MS then transmits the soft-coded rate bits to
9 the BS with the following frame.

10 Flow chart 1000 begins with first step 1002 which includes reception of
11 a radio-frequency frame at the BS. Following receipt of the frame at the BS,
12 the soft-coded rate bits are extracted from the frame data in step 1004. In a
13 preferred embodiment of the present invention and as discussed above in
14 conjunction with Figure 4, these soft-coded rate bits may include three (3) bits
15 of rate information that can identify up to eight (8) different vocoder and
16 channel coding rates. The frame data is then convolutionally decoded to yield
17 the original speech block in step 1006.

18 Using the appropriate vocoder and channel coding rate information
19 extracted in step 1004, the speech block is then decoded to recreate the
20 original voice data in step 1008. In this manner, the BS may receive a frame
21 containing voice data encoded with virtually any vocoder rate, and the frame
22 may be successfully decoded to the original voice data because all relevant

1 vocoder rate information is transmitted within the frame in the form of soft-
2 coded bits.

3 In order to provide the best possible voice communication channel, the
4 BS determines the channel quality metrics discussed above, such as FE,
5 BER and RXQ, in step 1010. The BS also calculates the SRXQ value in step
6 1012 and, based upon the results of the measured and calculated metrics,
7 determines the vocoder rate for optimal voice quality in step 1014. In a
8 preferred embodiment of the present invention, the rate bits corresponding to
9 the new vocoder and channel coding rate are determined from a look-up
10 table. Once the vocoder and channel coding rate is determined, the BS
11 transmits a frame with the new uplink vocoder rate convolutionally coded into
12 the frame in step 1016. Downlink 1018 represents the transmission of a
13 frame from the BS to the MS.

14 In step 1020, the MS receives the frame containing the
15 convolutionally-coded uplink vocoder rate for the next downlink transmission.
16 Because it is not necessary to know the uplink vocoder rate in order to
17 decode the uplink transmission, the uplink vocoder rate may be
18 convolutionally encoded instead of soft-coded.

19 In step 1022, the MS decodes the received frame from the BS yielding
20 the new uplink vocoder rate bits. These vocoder rate bits are used to
21 determine, using a look-up table or the like, the new uplink vocoder rate.
22 Using that newly determined vocoder rate, the MS encodes the voice data in

1 step 1024 for transmission to the BS. In step 1026, the MS transmits the
2 frame containing the convolutionally encoded voice data and soft-coded
3 uplink vocoder rate bits to the BS. Uplink 1028 represents the transmission
4 of a frame from the MS to the BS.

5 Importantly, each uplink message includes as soft-coded bits the rate
6 information related to the speech block. This soft-coding enables the BS to
7 properly decode the speech block without knowing in advance the vocoder
8 rate. This is particularly advantageous because there exists a possibility that
9 a frame may become corrupted and no longer readable. This corruption may
10 create a situation wherein the BS may have transmitted a message frame in
11 the downlink changing the uplink vocoder rate, and that frame was not
12 successfully received by the MS. If this occurs, the BS would expect to
13 receive a frame having a new vocoder rate, while the frame actually received
14 would be encoded at the old rate. Additionally, in circumstances involving
15 discontinuous transmissions (DTX), such as when the BS is not continuously
16 transmitting, the channel characteristics and corresponding vocoder rate
17 information could change significantly between transmitted frames.
18 Consequently, in order to avoid such mis-communication, each speech block
19 is soft-coded with the rate information necessary to decode the speech block.

20 In a preferred embodiment of the present invention as shown in Figure
21 10, steps within sequence 1000 identified by bracket 1030 are performed
22 within the BS, and steps within sequence 1000 identified by bracket 1032 are
23 performed within the MS.

1

2 System Performance

3 The communication system of the present invention provides for the
4 block and bit interleaving thereby minimizing the disruption to the
5 communication link caused by channel noise, interference, and dropped
6 frames. In addition to such redundancy, the vocoder rate information which is
7 either hard-coded within the frame or soft-coded outside the frame, may also
8 be repetitive. Such repetition will further enhance the resilience of the
9 communication system of the present invention. Redundancy of the vocoder
10 rate information, or rate bits, may be accomplished by repeating the bits in
11 several locations within the speech frame, as mentioned above in conjunction
12 with Figure 4.

13 Like traditional GSM-based communication systems, the
14 communication system of the present invention provides for the transfer, or
15 "hand-off," of a MS from one BS to another BS in a different cell. In such a
16 hand-off, it would not be necessary to provide the new BS with any special
17 rate information via the communication link 108 as all necessary vocoder rate
18 information is presented in each frame transmitted from the MS.

19 The present invention may be implemented in either a full-rate or half-
20 rate GSM-based communication system. The encoding and transmission of
21 the vocoder rate information between the BS and MS in both the full and half-
22 rate system would be identical.

1 In addition to the modification of the vocoder rate and channel coding
2 as discussed above, the power level of the transmissions may also be
3 modified in order to provide the best possible voice quality. For example, in
4 Figures 8 and 9, rate bits 806 and 902 may take into consideration, in
5 addition to the FE and RXQ' metrics, a metric related to the power level of the
6 transmission. In such a situation, the BS may adjust the vocoder rate and
7 channel coding, while at the same time adjusting the BS transmit power to
8 minimize the BER or FE, resulting in better voice quality.

9 While the present invention has been discussed at length with respect
10 to the transmission of voice data between a BS and a MS, it should be
11 appreciated that any digital data may be communicated in a similar manner.
12 In fact, because other types of digital data may not be dependent upon the
13 audio sampling rates, a much higher data rate may be achieved using the
14 present invention. and is fully contemplated herein.

15 OTHER EMBODIMENTS

16 While there have been shown what are presently considered to be
17 preferred embodiments of the invention, it will be apparent to those skilled in
18 the art that various changes and modifications can be made herein without
19 departing from the scope and spirit of the invention as defined by the
20 appended claims and their equivalents.

CLAIMS

What is claimed is:

- 1 1. A wireless communication system comprising:
2 a mobile station; and
3 a base station which transmits a downlink signal to the mobile station
4 wherein the downlink signal includes a downlink rate and
5 wherein the mobile station extracts the downlink rate from the
6 downlink signal and decodes the downlink signal using this
7 extracted downlink rate.
- 1 2. The wireless communication system of claim 1, wherein the
2 downlink rate is determined by the mobile station using one or more channel
3 quality metrics.
- 1 3. The wireless communication system of claim 2, wherein the
2 downlink rate is communicated to the base station by the mobile station
3 using one or more rate bits in a uplink signal.
- 1 4. The wireless communication system of claim 3, wherein the one or
2 more rate bits are soft-coded into the uplink signal.
- 1 5. The wireless communication system of claim 2, wherein
2 determination of the downlink rate comprises comparing one of the one or
3 more channel quality metrics to a predefined value, and selecting a downlink

4 rate corresponding to the predefined value.

1 6. The wireless communication system of claim 5, wherein the
2 downlink rate is communicated from the base station to the mobile station
3 using one or more rate bits in the downlink signal.

1 7. The wireless communication system of claim 6, wherein the one or
2 more rate bits are soft-coded into the downlink signal.

1 8. The wireless communication system of claim 2, wherein the one or
2 more channel quality metrics includes a frame erase metric.

1 9. The wireless communication system of claim 2, wherein the one or
2 more channel quality metrics includes a receive quality metric.

1 10. The wireless communication system of claim 2, wherein the one or
2 more channel quality metrics includes a bit-error-rate metric.

1 11. The wireless communication system of claim 2, wherein the one or
2 more channel quality metrics includes an average receive quality metric.

1 12. A wireless communication system comprising:
2 a mobile station which monitors one or more downlink channel quality
3 metrics, wherein the mobile station determines one or more
4 quantization bits corresponding to the one or more downlink
5 channel quality metrics; and

6 a base station which receives an uplink signal including the one or
7 more quantization bits from the mobile station, and wherein the
8 base station determines a downlink rate corresponding to the
9 one or more
10 quantization bits and transmits a downlink signal to the mobile
11 station using the downlink rate.

1 13. The wireless communication system of claim 12, wherein the
2 downlink signal includes the downlink rate.

1 14. The wireless communication system of claim 12, wherein
2 determination of the downlink rate comprises comparing one of the one or
3 more channel quality metrics to a predefined value, and selecting a downlink
4 rate corresponding to the predefined value.

1 15. The wireless communication system of claim 12, wherein the
2 downlink rate is communicated from the base station to the mobile station
3 using one or more rate bits in the downlink signal.

1 16. The wireless communication system of claim 15, wherein the one
2 or more rate bits are soft-coded into the downlink signal.

1 17. The wireless communication system of claim 12, wherein the one
2 or more channel quality metrics includes a frame erase metric.

1 18. The wireless communication system of claim 12, wherein the one
2 or more channel quality metrics includes a receive quality metric.

1 19. The wireless communication system of claim 12, wherein the one
2 or more channel quality metrics includes a bit-error-rate metric.

1 20. The wireless communication system of claim 12, wherein the one
2 or more channel quality metrics includes an average receive quality metric.

1 21. A wireless communication system comprising:
2 a base station; and
3 a mobile station which transmits an uplink signal to the base station,
4 wherein the uplink signal includes an uplink rate and wherein
5 the base station
6 extracts the uplink rate from the uplink signal and decodes the
7 uplink signal using this extracted uplink rate.

1 22. The wireless communication system of claim 21, wherein the
2 uplink rate is determined in the base station and communicated to the mobile
3 station.

1 23. The wireless communication system of claim 21, wherein the
2 uplink rate information is determined by the base station using one or more
3 channel quality metrics.

1 24. The wireless communication system of claim 23, wherein the
2 channel quality metrics includes a frame erase metric.

1 25. The wireless communication system of claim 23, wherein the
2 channel quality metrics includes a receive quality metric.

1 26. The wireless communication system of claim 23, wherein the
2 channel quality metrics includes a bit-error-rate metric.

1 27. The wireless communication system of claim 23, wherein the
2 channel quality metrics includes an average receive quality metric.

1 28. A method of maximizing voice quality in a wireless communication
2 system including a base station and a mobile station, the method comprising
3 the steps of:
4 determining the quality of a downlink communication channel in the
5 mobile station using one or more channel quality metrics;
6 determining a new downlink rate based on the channel quality metrics;
7 transmitting the new downlink rate from the mobile station to the base
8 station
9 in an uplink signal;
10 extracting the new downlink rate from the uplink signal;
11 creating a downlink signal using the new downlink rate;
12 encoding the new downlink rate into the downlink signal;

13 transmitting the downlink signal from the base station to the mobile
14 station;
15 receiving the downlink signal at the mobile station;
16 extracting the new downlink rate from the downlink signal; and
17 decoding the downlink signal using the extracted new downlink rate.

1 29. A method of maximizing voice quality in a wireless communication
2 system including a base station and a mobile station, the method comprising
3 the steps of:
4 determining the quality of a downlink communication channel in the
5 mobile station using one or more channel quality metrics;
6 selecting one or more quantization bits corresponding to the one or
7 more
8 channel quality metrics from a predetermined group of
9 quantization bits;
10 transmitting the quantization bits from the mobile station to the base
11 station in an uplink signal;
12 extracting the quantization bits from the uplink signal in the base
13 station;
14 determining a downlink rate from the extracted quantization bits;
15 creating a downlink signal using the downlink rate;
16 encoding the downlink rate into the downlink signal;
17 transmitting the downlink signal from the base station to the mobile

18 station;
19 receiving the downlink signal at the mobile station;
20 extracting the downlink rate from the downlink signal; and
21 decoding the downlink signal using the extracted downlink rate.

1 30. A method of maximizing voice quality in a wireless communication
2 system including a base station and a mobile station, the method comprising
3 the steps of:
4 determining the quality of an uplink communication channel in the
5 base
6 station using one or more channel quality metrics;
7 determining an uplink rate based on the channel quality metrics;
8 transmitting the uplink rate from the base station to the mobile station
9 in a downlink signal;
10 extracting the uplink rate from the downlink signal;
11 creating an uplink signal using the uplink rate;
12 encoding the new uplink rate into the uplink signal;
13 transmitting the uplink signal from the mobile station to the base
14 station;
15 receiving the uplink signal at the base station;
16 extracting the new uplink rate from the uplink signal; and
17 decoding the uplink signal using the extracted uplink rate.

1 31. A wireless communication system comprising:
2 means for determining the quality of a downlink communication
3 channel in the mobile station using one or more channel quality
4 metrics;
5 means for determining a new downlink rate based on the channel
6 quality metrics;
7 means for transmitting the new downlink rate from the mobile station to
8 the
9 base station in an uplink signal;
10 means for extracting the new downlink rate from the uplink signal;
11 means for creating a downlink signal using the new downlink rate;
12 means for encoding the new downlink rate into the downlink signal;
13 means for transmitting the downlink signal from the base station to the
14 mobile station;
15 means for receiving the downlink signal at the mobile station;
16 means for extracting the new downlink rate from the downlink signal;
17 and
18 means for decoding the downlink signal using the extracted new
19 downlink rate.

1 32. A wireless communication system comprising:
2 means for determining the quality of a downlink communication
3 channel in

4 the mobile station using one or more channel quality metrics;
5 means for selecting one or more quantization bits corresponding to the
6 one or more channel quality metrics from a predetermined
7 group of
8 quantization bits;
9 means for transmitting the quantization bits from the mobile station to
10 the
11 base station in an uplink signal;
12 means for extracting the quantization bits from the uplink signal in the
13 base station;
14 means for determining a downlink rate from the extracted quantization
15 bits;
16 means for creating a downlink signal using the downlink rate;
17 means for encoding the downlink rate into the downlink signal;
18 means for transmitting the downlink signal from the base station to the
19 mobile station;
20 means for receiving the downlink signal at the mobile station;
21 means for extracting the downlink rate from the downlink signal; and
22 means for decoding the downlink signal using the extracted downlink
23 rate.

1 33. A wireless communication system comprising:
2 means for determining the quality of an uplink communication channel

3 in the base station using one or more channel quality metrics;
4 means for determining an uplink rate based on the channel quality
5 metrics;
6 means for transmitting the uplink rate from the base station to the
7 mobile station in a downlink signal;
8 means for extracting the uplink rate from the downlink signal;
9 means for creating an uplink signal using the uplink rate;
10 means for encoding the new uplink rate into the uplink signal;
11 means for transmitting the uplink signal from the mobile station to the
12 base station;
13 means for receiving the uplink signal at the base station;
14 means for extracting the new uplink rate from the uplink signal; and
15 means for decoding the uplink signal using the extracted uplink rate.

1 34. A wireless communication system, comprising:
2 a base station;
3 a mobile station which receives a downlink signal from the base station
4 over a downlink, the downlink signal having a ratio of voice data
5 to channel coding;
6 means for determining the quality of the downlink; and
7 means for adjusting the ratio of the voice data to channel coding of the
8 downlink signal to improve the quality of the downlink.

1 35. The wireless communication system of claim 34, wherein the
2 means for determining the quality of the downlink comprises one or more
3 channel quality metrics.

1 36. A wireless communication system, comprising:
2 a mobile station;
3 a base station which receives an uplink signal from the mobile station
4 over an uplink, the uplink signal having a ratio of voice data to
5 channel coding;
6 means for determining the quality of the uplink; and
7 means for adjusting the ratio of the voice data to channel coding of the
8 uplink signal to improve the quality of the uplink.

1 37. The wireless communication system of claim 36, wherein the
2 means for determining the quality of the uplink comprises one or more
3 channel quality metrics.

1/9

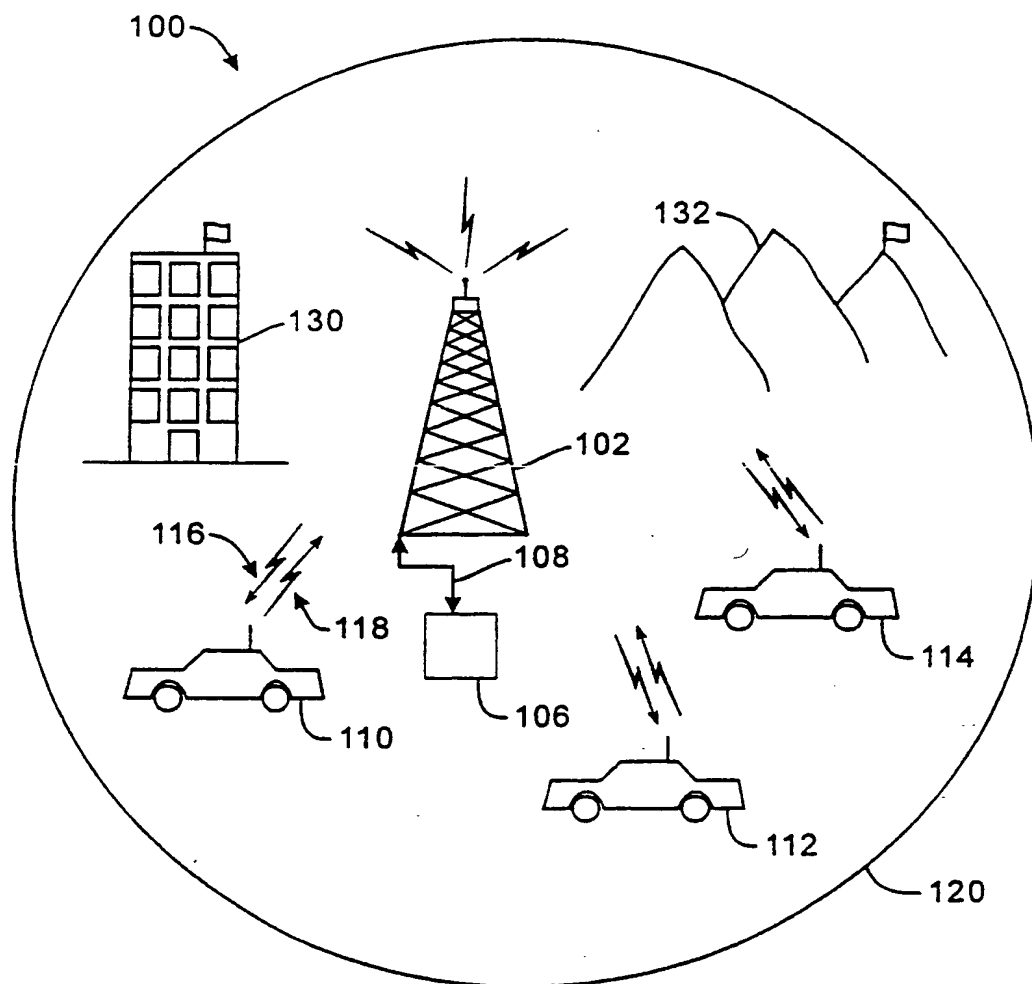


Fig. 1

SUBSTITUTE SHEET (RULE 26)

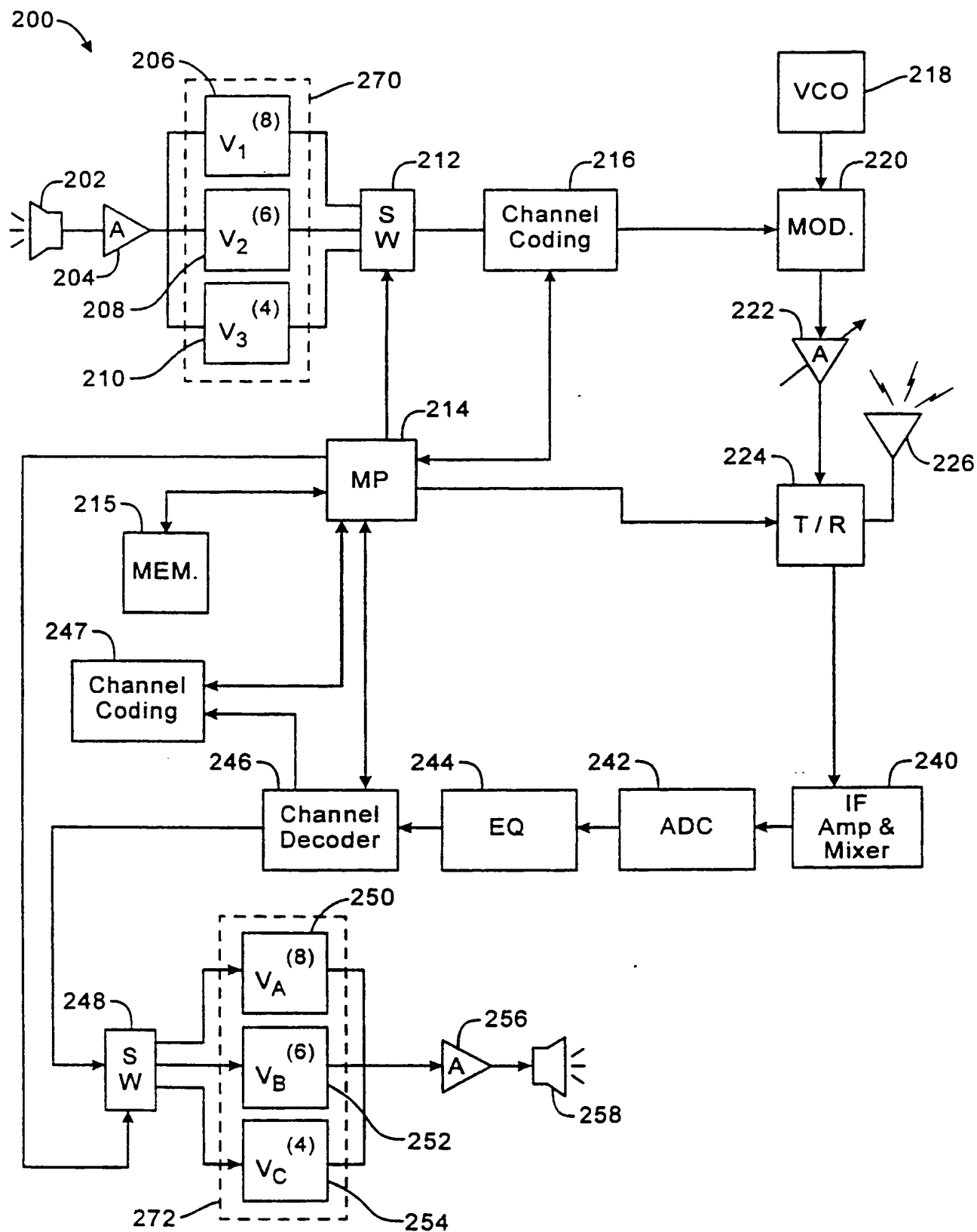


Fig. 2

SUBSTITUTE SHEET (RULE 26)

3/9

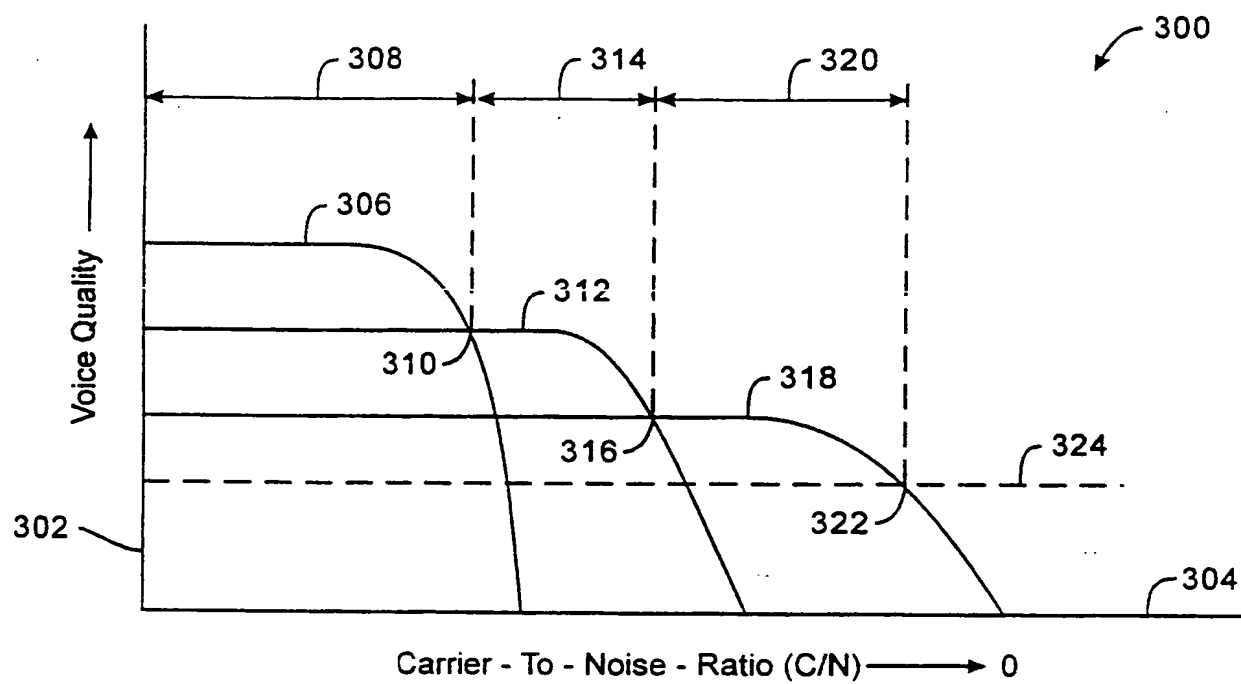


Fig.3

4/9

400

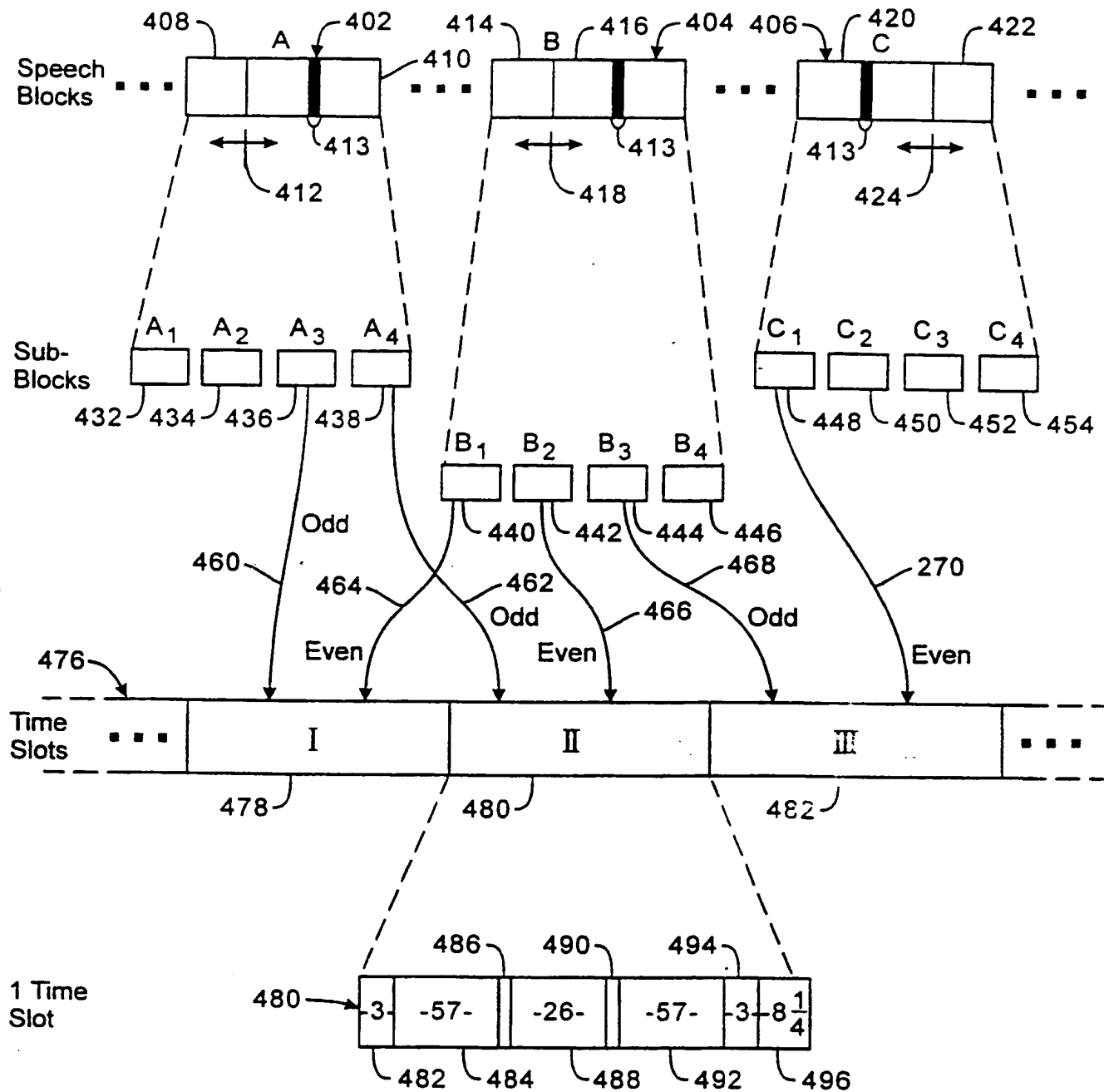


Fig. 4

SUBSTITUTE SHEET (RULE 26)

5/9

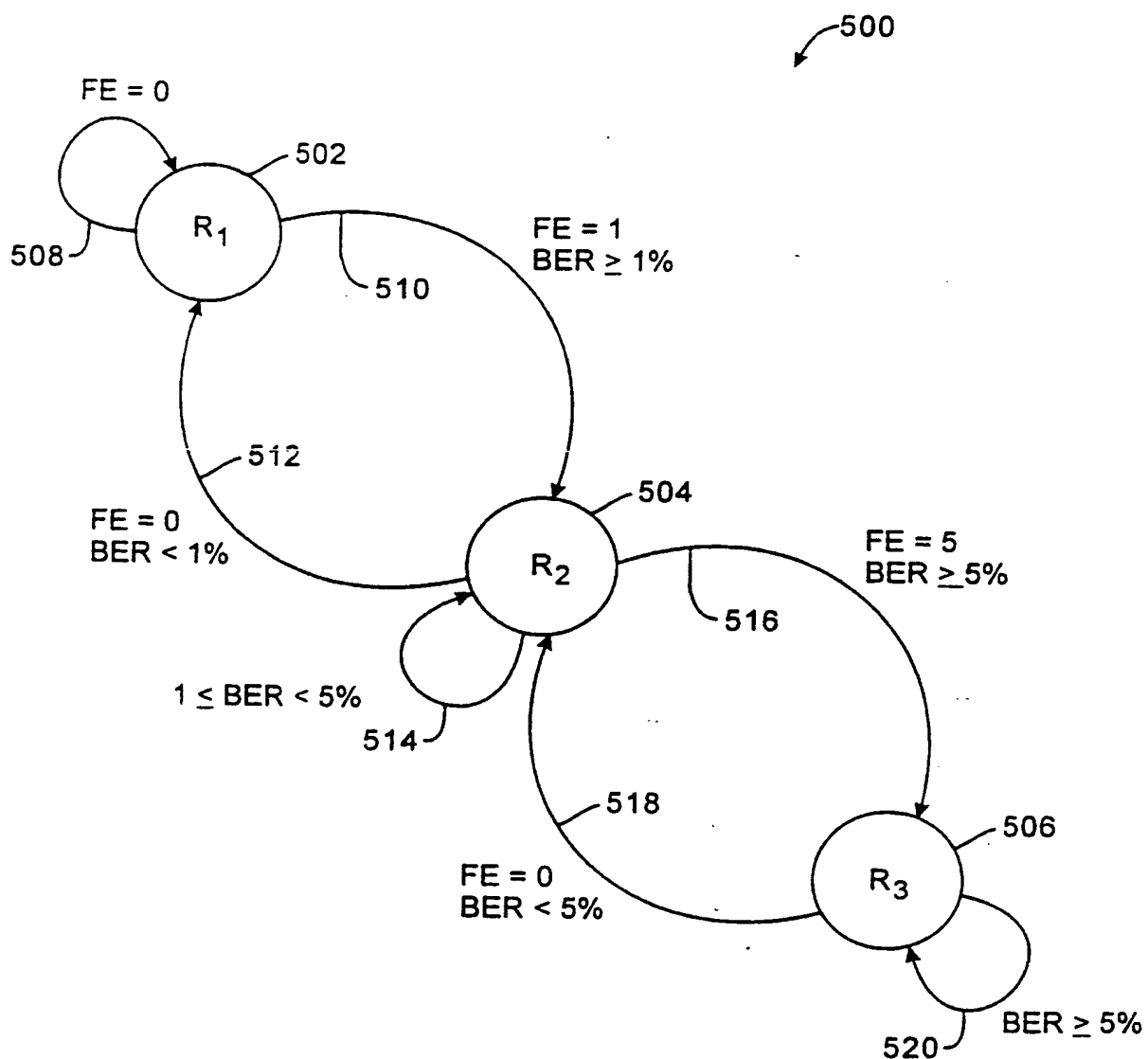


Fig. 5

SUBSTITUTE SHEET (RULE 26)

6/9

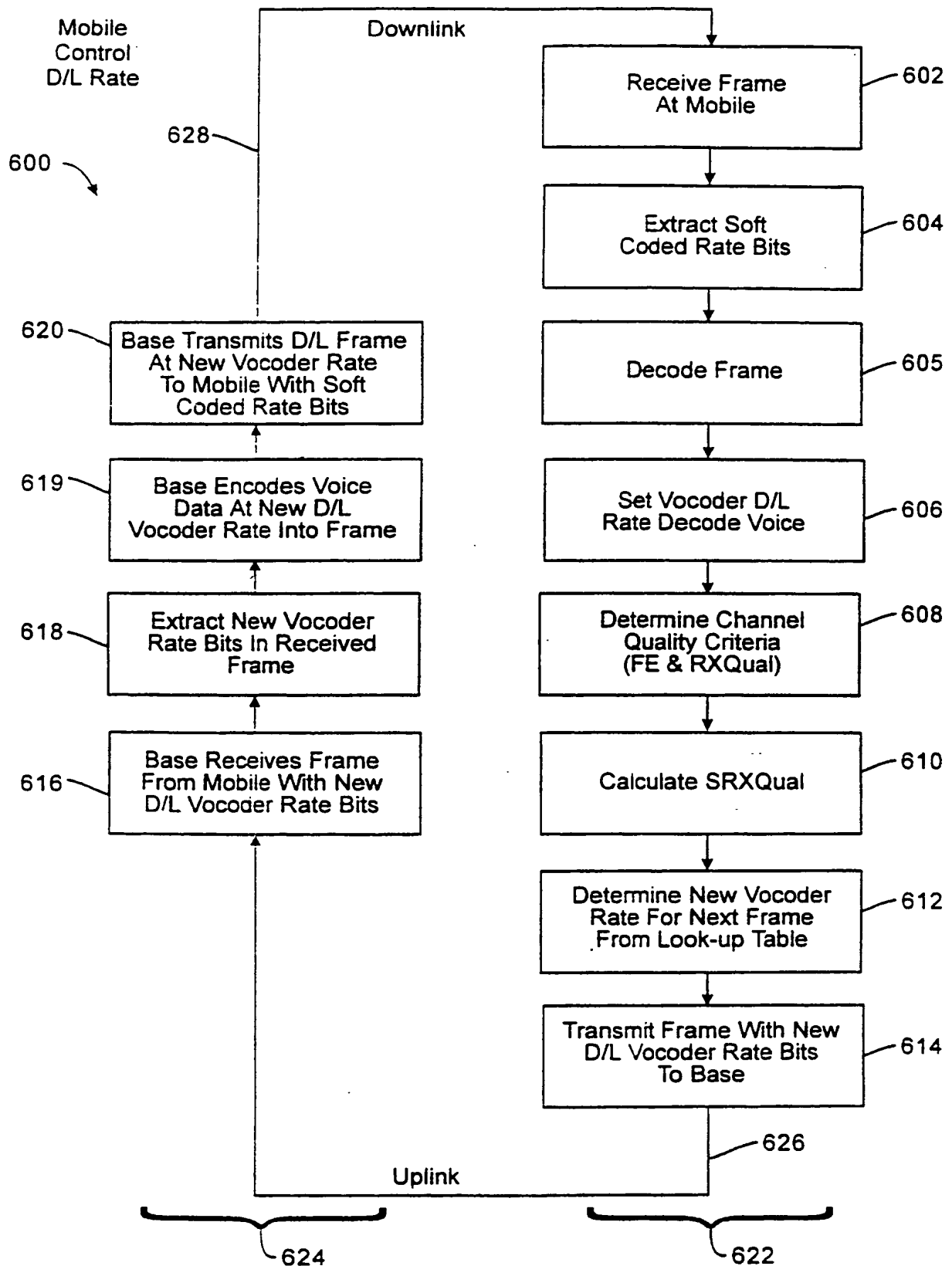


Fig. 6

SUBSTITUTE SHEET (RULE 26)

7/9

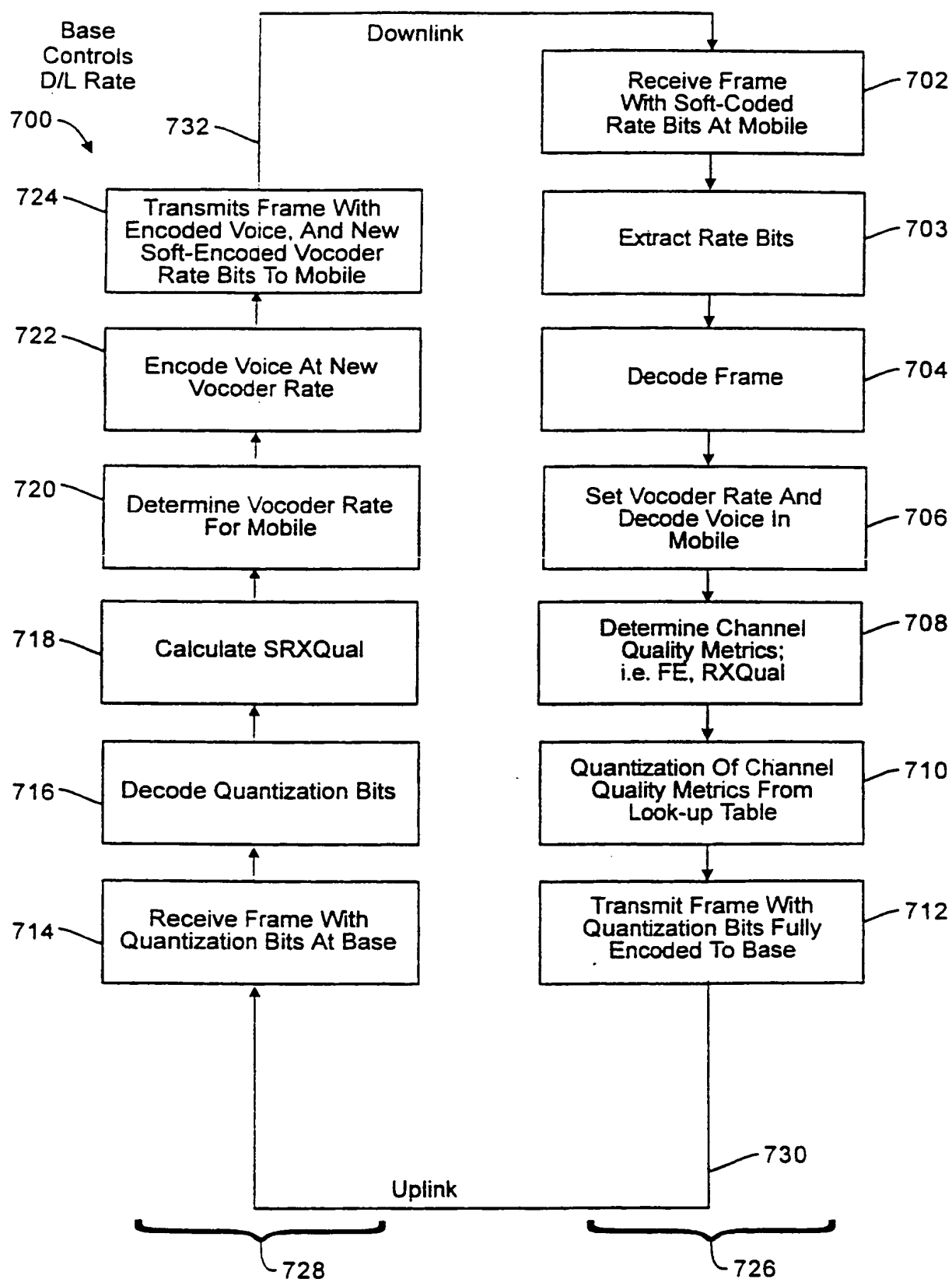


Fig. 7

SUBSTITUTE SHEET (RULE 26)

8/9

(Quantization Table For The Mobile)

Frame Erase (FE)	RXQual < = 804	Transmitted Bits
0	5	000
0	10	001
0	15	010
0	20	011
1	25	100
1	35	101
1	60	110
1	80	111

Fig. 8

(Quantization Table For The Base Station)

Received Bits	Frame Eraser (FE) 904	RXQual
000	0	2
001	0	7
010	0	12
011	0	17
100	1	22
101	1	29
110	1	37
111	1	65

Fig. 9

SUBSTITUTE SHEET (RULE 26)

9/9

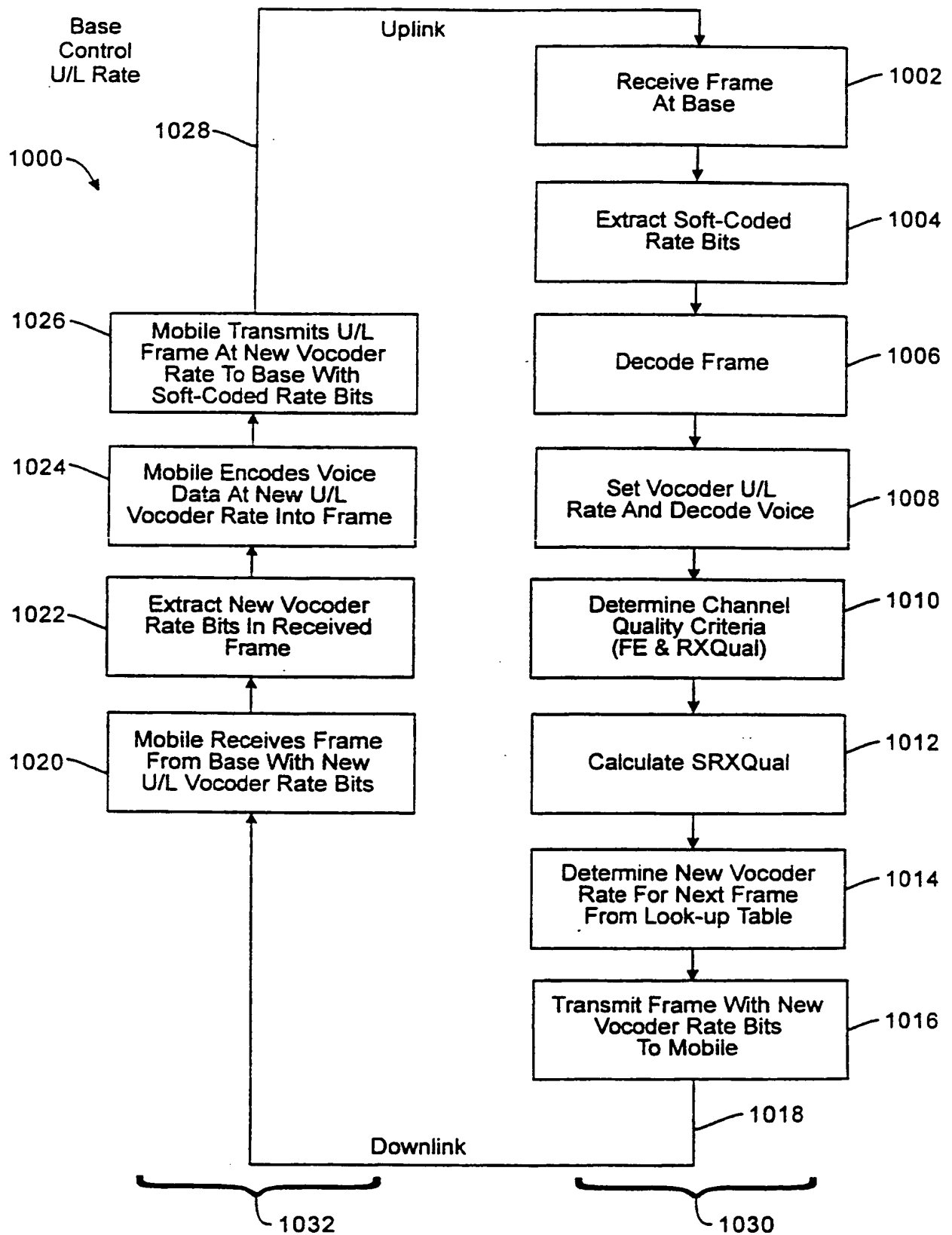


Fig. 10

SUBSTITUTE SHEET (RULE 26)

INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 99/10775

A. CLASSIFICATION OF SUBJECT MATTER
IPC 6 H04L1/12

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	WO 98 03030 A (NOKIA MOBILE PHONES LTD ;JOKINEN HARRI (FI); HAKASTE MARKUS (FI);) 22 January 1998 (1998-01-22) abstract page 3, line 4 - line 12 page 3, line 21 - line 25 page 5, line 4 - line 16 page 5, line 36 - line 37 page 6, line 21 - line 26 page 7, line 31 - page 8, line 4 figures 1-4	1-3, 5, 6, 9, 10, 12-15, 18, 19, 21-23, 25, 26, 28, 30, 31, 33-37
Y	-/--	8, 11, 17, 20, 24, 27, 29, 32

☒ Further documents are listed in the continuation of box C.☒ Patent family members are listed in annex.

* Special categories of cited documents :

- "A" document defining the general state of the art which is not considered to be of particular relevance
- "E" earlier document but published on or after the international filing date
- "L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- "O" document referring to an oral disclosure, use, exhibition or other means
- "P" document published prior to the international filing date but later than the priority date claimed

- "T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
- "X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- "Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.
- "&" document member of the same patent family

Date of the actual completion of the international search

13 September 1999

Date of mailing of the international search report

21/09/1999

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Langinieux, F

INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 99/10775

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication where appropriate, of the relevant passages	Relevant to claim No.
X	<p>page 11, line 11 - line 17</p> <p>FR 2 718 906 A (ALCATEL MOBILE COMM FRANCE) 20 October 1995 (1995-10-20)</p> <p>abstract</p> <p>page 5, line 21 - line 26</p> <p>page 6, line 5 - line 14</p> <p>page 8, line 17 - line 22</p> <p>page 10, line 15 - line 23</p> <p>page 11, line 16 - line 19</p> <p>page 15, line 21 - line 24</p> <p>page 16, line 8 - line 10</p> <p>page 24, line 8 - line 12</p>	<p>1-3,5,6, 10, 12-15, 19, 21-23, 26,28, 30,31, 33-37</p>
Y	<p>WO 96 22639 A (QUALCOMM INC)</p> <p>25 July 1996 (1996-07-25)</p> <p>page 20, line 37 - page 21, line 8</p>	<p>8,17,24</p>
Y	<p>GB 2 306 867 A (BOSCH GMBH ROBERT)</p> <p>7 May 1997 (1997-05-07)</p> <p>abstract</p> <p>page 2, line 2 - line 9</p> <p>page 2, line 26 - line 30</p> <p>page 3, line 12 - line 21</p> <p>page 4, line 1 - line 18</p> <p>figure 2</p>	<p>11,20, 27,29,32</p>

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PL/US 99/10775

Patent document cited in search report		Publication date	Patent family member(s)	Publication date
WO 9803030	A	22-01-1998	FI 962834 A	13-01-1998
			AU 3445397 A	09-02-1998
FR 2718906	A	20-10-1995	AU 697394 B	01-10-1998
			AU 2311295 A	10-11-1995
			AU 9827498 A	04-03-1999
			CA 2187669 A	26-10-1995
			EP 0755615 A	29-01-1997
			FI 964043 A	04-12-1996
			WO 9528814 A	26-10-1995
			JP 9512672 T	16-12-1997
			NZ 284502 A	20-12-1996
WO 9622639	A	25-07-1996	US 5568483 A	22-10-1996
			AU 694612 B	23-07-1998
			AU 4760396 A	07-08-1996
			BR 9606833 A	30-12-1997
			CA 2210657 A	25-07-1996
			EP 0804836 A	05-11-1997
			FI 972990 A	17-09-1997
			JP 10512415 T	24-11-1998
			ZA 9600181 A	14-10-1996
GB 2306867	A	07-05-1997	DE 19603725 A	30-04-1997
			JP 9149009 A	06-06-1997
			US 5926232 A	20-07-1999